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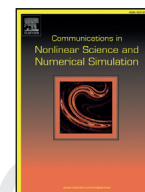
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Research paper

# Size-dependent nonlinear forced vibration and dynamic stability of electrically actuated micro-plates

I. Karimipour<sup>a,b</sup>, Y. Tadi Beni<sup>a,\*</sup>, A.H. Akbarzadeh<sup>a,c</sup><sup>a</sup> Department of Mechanical Engineering, Shahrood University, Shahrood, Iran<sup>b</sup> AML Laboratory, Department of Bioresource Engineering, McGill University, Island of Montreal, QC H9X3V9 Canada<sup>c</sup> Department of Mechanical Engineering, McGill University, Montreal, QC H3A 0C3 Canada

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## ABSTRACT

A size-dependent structural dynamic model that incorporates the effect of geometric non-linearity is developed in this paper for the forced vibration, and dynamic stability of thin rectangular micro-plates. The equations of motion for micro-plates are derived within the framework of classical plate theory, modified couple stress theory (MCST), and von Kármán geometric nonlinearity using Hamilton's principle. Galerkin method is used to convert the governing partial differential equations to a nonlinear second-order ordinary differential equation, which is solved by a Runge-Kutta method. The static instability analysis of the micro-plate is performed to determine the critical electrostatic voltages, and to avoid the pull-in instability. By tracking the static behavior of the microplate, and determining the electrostatic pull-in voltage, the frequency response curves are plotted. In dynamic response, primary, superharmonic, and subharmonic resonance are studied, and the frequency response equation is obtained for each case by the method of multiple scales. Further efforts are made to investigate the influence of size effect, electrical loading (DC and AC voltages), and excitation frequency on the static, and dynamic responses, critical AC voltages, and dynamic stability of micro-plates. It is found that the critical dynamic voltage is a function of the frequency of excitation force. It is shown that the stiffness of micro-plate decreases by increasing the constant DC voltage; however, the increase in the alternating AC voltage does not considerably affect the stiffness of the micro-plate.

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## 1. Introduction

Recent studies show that electrically-actuated micro-plates have a vast range of applications in microelectromechanical systems (MEMS) as actuation components in micro-pumps, micro-mirrors, microphones, micro-switches, and micro-sensors [1–6]. The actuation force can be applied to the micro-plates in the form of electrical, magnetic and thermal excitation, so-called as multi-physical stimuli [3,7–10]. In the case of electrical actuation, an electrically-actuated micro-plate is formed by a variable capacity air-gap capacitor on one side and a stationary electrode connected to the output circuit on the other side [11,12]. Electrostatic actuators contain two conductive electrodes, one movable and one fixed (grounded). Applying a voltage between the electrodes leads to the deflection of the movable electrode toward the fixed electrode [13,14]. Many studies in the literature, however, have overlooked the bending stiffness and modelled the plate as a membrane element [1,11,15–19].

\* Corresponding author.

E-mail address: [tadi@eng.sku.ac.ir](mailto:tadi@eng.sku.ac.ir) (Y.T. Beni).<https://doi.org/10.1016/j.cnsns.2019.104856>

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Many researchers have investigated the behavior of electrically-stimulated micro-beams in resonator sensors. These studies have fallen into two categories. The first category focused on the static behavior of micro-beams bent by a DC electrostatic force. The second category studied the vibration behavior of micro-beams by a harmonic AC voltage. Zook et al. [20] considered micro-plates and micro-beams and calculated their fundamental frequencies using a finite element method. The natural frequencies obtained by the finite element method were higher than the measured natural frequencies measured by experimental tests. Choi and Lovell [21] calculated the static deformation of a micro-beam using a numerical integration scheme and a shooting method. Their models included electrostatic force and mid-plane stretching. Ahn et al. [22] modeled an electrically-driven micro-beam as one degree of freedom mass-spring-damper system. They used this model to obtain an analytical expression for the fundamental natural frequency as a function of constant DC voltage. The second category of research was initiated by Zook et al. [20]. They showed that increasing the AC voltage increases the resonance frequency (hardening behavior). Tilmans and Legtenberg [23] studied the dynamic problem of the system for the large values of the AC voltage using the Rayleigh-Ritz energy method, taking into account the electric forces and the stretchability of the mid-plane. Ayela and Fournier [24] studied the response of micro-beams with different geometric shapes under a total electrical excitation including DC and AC voltages. They plotted different diagrams to show the variation of resonant frequency relative to the excitation amplitude for different axial loads. Laboratory results showed that some of the micromachined silicon resonators have a softening behavior for the operating conditions. While others have a hardening behavior, they concluded that a nonlinear behavior might be due to different phenomena, e.g., the mechanical properties of these resonators. Veijola et al. [25] modeled a micro-beam using a nonlinear mass and spring model of order three, which includes the stretch of the intermediate plane. Using the Harmonic Balance Method, they showed that a nonlinear term caused by electrical force leads to a softening behavior, while the stretching of the mid-plane results in a hardening behavior.

Pull-in instability in micro-electromechanical devices is investigated by König and Wachutka [26]. This instability occurs when the input voltage trespasses a critical value called the pull-in voltage [27]. In this way, the elastic restoring force of the movable electrode cannot resist the Coulomb attraction force, and this electrode abruptly adheres to the fixed one. Ng et al. [28] have studied the characteristics of an electrically stimulated plate. The governing Laplace equation is solved by the boundary element method. Also, the nonlinear geometric factors related to the tension of the mid-plane are included in the plate model. Subsequently, electromechanical coupling equations were solved by the repetition method. Significant qualitative differences were observed between the results of the linear and nonlinear analysis for large plate deformations. Zhao et al. have studied nonlinear modeling of simply supported rectangular plates [29]. They determined static deformation using a numerical shooting method and a reduced order method. They also studied the mechanical behavior of rectangular plates under electrical excitation. The linear and nonlinear vibration of plates have been studied by a large number of researchers. Analytical methods, along with numerous numerical theories, have been widely used in practice [30].

After the dimensions have reduced to a sub-micron scale, the nano-scale phenomena emerge, which should be taken into account in establishing the theoretical models [31,32]. One of the major critical instances of the nano-scale phenomenon is the size-dependency of the mechanical performance of nanostructures, which appears in the deformation tests of microstructures [33]. Due to the fact that the materials at the atomic scale are naturally discrete, the classical continuum mechanics are supposed to be insufficiently effective for modeling the size-dependent behavior of them at sub-micron distances. The ever-increasing progress in micro-structure materials leads to extending the usages of higher-order continuum theories. In constructing a large number of devices in a nano-scale, the classical elasticity, due to neglect size effect, loses its efficiency [34]. In the past decades, the researchers and experts have extensively and particularly adopted higher-order continuum theories in the nano-scale studies of thin films, nano-composites, and quantum dots. Employing the classical continuum theory in the problems which encompass thin films, nano-composites, and quantum dots, yielded extremely unexpected results [35]. Couple stress theory [36], non-local elasticity theory [37], micro-polar elasticity theory [38], strain gradient elasticity theory [39–42] and surface elasticity [43] are good instances of the theories developed and used to study the mechanical behaviors of micro-scale structures [31]. Modified couple stress theory introduces one material length scale parameter as an additional elastic constant to interpret the size-dependent behavior of elastic solids [44]. In the following, some of the works on the modified couple stress theory will be reviewed. Tsiatas [45] presented a size-dependent model for testing the static flexure of thin micro-plates based on assumptions Kirchhoff model. He concluded that in the smaller thickness of the micro-plate or the more significant amount of its material length to its thickness ( $l_h$ ), the influence of size effect increases. Using the method presented by Tsiatas, Yin et al. [46] introduced the non-classic model of Kirchhoff's plate to investigate the effect of size on the first two natural frequencies of micro-plates based on modified couple stress theory. Also, Jomehzadeh et al. [47] using the model provided by Tsiatas, analytically investigated the effect of size on the natural frequency of thin simply supported micro-plates based on modified couple stress theory and Kirchhoff assumptions. They also examined circular plates with different boundary conditions. They showed that by decreasing the thickness of the micro-plates, the value of natural frequency significantly increased. Then Asghari [48] expanded Tsiatas's work by considering non-linear geometric effects in equations. He also presented the size-dependent model for thin plates by recognizing nonlinear geometric effects based on the modified couple stress theory. Wang et al. [49] presented a non-classical model for Kirchhoff plates based on the principle of minimum potential energy to analyze nonlinear bending micro-circular plates under a uniform load. The governing equations first were converted to nonlinear algebraic equations using Collocation point method, and then these equations were solved using the Newton- Raphson numerical method. Numerical results showed the plate that is modeled using MCST is stiffer than the plate is modeled using classical theory, so the lower ratio of thickness

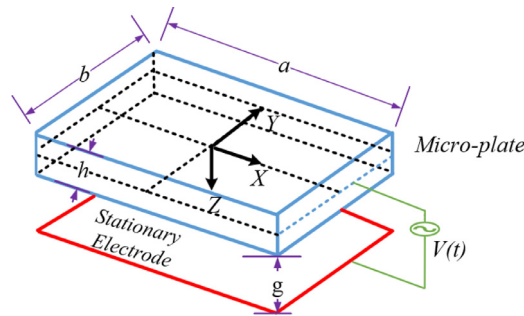


Fig. 1. Schematic figure of a rectangular micro-plate.

to the material length scale parameter, the difference between two theories gets bigger. Akgoz and Civalek [50] investigated an analytical solution for static bending, and free vibration of micro-plates rested on an elastic foundation based on modified couple stress theory. Equations were derived based on the Kirchhoff plate theory, and for solving equations, the Navier's method was used. Askari and Tahani [51] extracted the size-dependent natural frequency of thin-rectangular micro-plate with clamped boundary condition based on the modified couple stress theory using the extended Kantorovich method (EKM). They also [52] investigated the size effects on the natural frequencies in thin clamped micro-plates under the electrostatic field. They concluded that accounting the size effects in the free vibration analysis of pre-deformed micro-plate, under applying the electrical potential is more urgent than one with un-deformed structure. Tahani et al. [53] investigated the effects of length scale on natural frequencies and linear and un-damped mode shapes of thin rectangular micro-plates. The micro-plate has initial deflection under the presence of an electrostatic field. They used the finite element method to solve equations and showed that the convergence of results is gained by using a  $20 \times 20$  grid of points. They concluded that the influence of the size effect on the natural frequencies of the pre-deformed plates when  $h > 20l$  (thickness ratio to a material length scale) is negligible, while considering the size effect is essential for  $h < 10l$ . Zhang et al. [54] presented the size-dependent finite element model for thick Mindlin micro-plates using modified couple stress theory. They studied static bending, buckling, and free vibration of thick micro-plates using the finite element model. It should be noted that all the mentioned papers have been examined for homogeneous micro-plates. Recently, the modified couple stress theory is used to analyze the functionally graded micro-plate. A literature survey on these topics has been carried out in [55–59].

Due to the review carried out in previous studies in the field of the micro-plates, the main objective of this work is study on the nonlinear forced vibration and dynamic behavior of rectangular micro-plates under electrical excitation. To fill this gap, the present study adopts MCST components together with developing a semi-analytical method which is appropriate for studying the micro-plate's forced vibration. The applied excitation is made up of a constant current voltage DC and a variable AC voltage. In this analysis, the first step is to achieve the equations of motion under the electrical excitation and to simplify them. Then, to obtain the static response of the system under constant voltage (DC voltage), these equations are solved by different methods. By identifying the critical voltage and applying electrical excitation consisting of constant and alternating current voltages, the frequency response curves for the primary and secondary resonance modes are investigated. Other objectives of this study are to study the effects of electrical load parameters, (namely DC and AC driving voltages), and the influence of excitation frequency on the static and vibrational behavior of the micro-plate. In this regard, semi-analytical relations for static and vibration responses of the micro-plate under electrical stimulation are obtained, and critical voltage and the dynamic stability of the system are reported.

## 2. Size-dependent micro-plate model

The classical plate theory (CPT) is applied to plates whose thickness is small compared to other dimensions, where the transverse shear deformation, rotation inertia, and normal transverse stress can be neglected [60]. Fig. 1 demonstrates a schematic figure of a micro-plate, constructed of two conductive electrodes (one movable and one fixed (grounded)). The considered micro-plate has the length and width of  $a$  and  $b$  in the  $X$  and  $Y$  directions, while its thickness is  $h$ . The initial gap between the non-actuated micro-plate and the fixed substrate is assumed to be  $g$ . In addition,  $X$ ,  $Y$ , and  $Z$  are the coordinates along the length, width, and thickness, respectively.

According to the underlying hypothesis of the classical thin micro-plate theory, the displacement field  $(U_1, V_1, W_1)$  of an arbitrary point of the micro-plate can be specified as [61]:

$$U_1(X, Y, Z, t) = U(X, Y, t) - Z \frac{\partial}{\partial X} W(X, Y, t),$$

$$V_1(X, Y, Z, t) = V(X, Y, t) - Z \frac{\partial}{\partial Y} W(X, Y, t),$$

$$W_1(X, Y, Z, t) = W(X, Y, t)$$

(1)

where  $(U_1, V_1, W_1)$  are the corresponding displacements along with the  $(X, Y, Z)$  coordinates. In Eq. (1),  $U, V$ , and  $W$  represent the mid-plane displacements along the coordinate directions.

Since the thickness of the micro-plate compared to other dimensions is assumed to be small, the classical plate theory is adopted in this research. The nonlinear analysis of the plates is of great importance when the amplitude of lateral vibration of the micro-plate is greater than the half of the thickness of the micro-plate, and therefore an essential nonlinear term should be considered in the governing equation. Due to the presence of a nonlinear geometric term, resonance frequencies and the mode shapes are dependent on the amplitude of lateral load. von Kármán equations which incorporate the effect of mid-plane stretch are widely used in the free and forced nonlinear vibration of plates [62]. Since later deformation is comparable to the thickness of micro-plates, classical plate theory is applied for a geometrically-nonlinear problem. For micro-plates with insignificant strains, moderate slopes, and large deflections, the non-zero strain components associated with the displacement field, based on the von Kármán equation, can be written as [61, 63, 64]:

$$\varepsilon_{ZZ} = \varepsilon_{XZ} = \varepsilon_{YZ} = 0,$$

$$\begin{Bmatrix} \varepsilon_{XX} \\ \varepsilon_{YY} \\ \gamma_{XY} \end{Bmatrix} = \begin{Bmatrix} \varepsilon_{XX}^0 \\ \varepsilon_{YY}^0 \\ \gamma_{XY}^0 \end{Bmatrix} - Z \begin{Bmatrix} \frac{\partial^2 W}{\partial X^2} \\ \frac{\partial^2 W}{\partial Y^2} \\ 2 \frac{\partial^2 W}{\partial X \partial Y} \end{Bmatrix}, \quad \begin{Bmatrix} \varepsilon_{XX}^{(0)} \\ \varepsilon_{YY}^{(0)} \\ \gamma_{XY}^{(0)} \end{Bmatrix} = \begin{Bmatrix} \frac{\partial U}{\partial X} + \frac{1}{2} \left( \frac{\partial W}{\partial X} \right)^2 \\ \frac{\partial V}{\partial Y} + \frac{1}{2} \left( \frac{\partial W}{\partial Y} \right)^2 \\ \frac{\partial U}{\partial Y} + \frac{\partial V}{\partial X} + \frac{\partial W}{\partial X} \frac{\partial W}{\partial Y} \end{Bmatrix} \quad (2)$$

The non-zero components of the rotation vector and curvature tensor, associated with the displacement field presented in Eq. (1), can also be written as:

$$\begin{Bmatrix} \theta_{XX} \\ \theta_{YY} \\ \theta_{ZZ} \end{Bmatrix} = \frac{1}{2} \text{Curl} \begin{Bmatrix} U_1 \\ V_1 \\ W_1 \end{Bmatrix} = \begin{Bmatrix} \frac{1}{2} \left( \frac{\partial W_1}{\partial Y} - \frac{\partial V_1}{\partial Z} \right) \\ \frac{1}{2} \left( \frac{\partial U_1}{\partial Z} - \frac{\partial W_1}{\partial X} \right) \\ \frac{1}{2} \left( \frac{\partial V_1}{\partial X} - \frac{\partial U_1}{\partial Y} \right) \end{Bmatrix}, \quad (3)$$

$$\begin{Bmatrix} \chi_{XX} \\ \chi_{YY} \\ \chi_{XY} \\ \chi_{XZ} \\ \chi_{YZ} \end{Bmatrix} = \begin{Bmatrix} \frac{\partial^2 W}{\partial X \partial Y} \\ -\frac{\partial^2 W}{\partial X \partial Y} \\ -\frac{1}{2} \left( \frac{\partial^2 W}{\partial X^2} - \frac{\partial^2 W}{\partial Y^2} \right) \\ -\frac{1}{4} \left( \frac{\partial^2 U}{\partial X \partial Y} - \frac{\partial^2 V}{\partial X^2} \right) \\ -\frac{1}{4} \left( \frac{\partial^2 U}{\partial Y^2} - \frac{\partial^2 V}{\partial X \partial Y} \right) \end{Bmatrix}, \quad (4)$$

A brief review on the Modified Couple Stress Theory (MCST) is given in Appendix A. According to Kirchhoff's hypothesis, the substitution of Eq. (2) into Eq. (A.2) results in the following stress components as functions of displacements:

$$\begin{Bmatrix} \sigma_{XX} \\ \sigma_{YY} \\ \sigma_{XY} \end{Bmatrix} = \frac{E}{1-\nu^2} \begin{bmatrix} 1 & \nu & 0 \\ \nu & 1 & 0 \\ 0 & 0 & (1-\nu)/2 \end{bmatrix} \left( \begin{Bmatrix} \frac{\partial U}{\partial X} + \frac{1}{2} \left( \frac{\partial W}{\partial X} \right)^2 \\ \frac{\partial V}{\partial Y} + \frac{1}{2} \left( \frac{\partial W}{\partial Y} \right)^2 \\ \frac{\partial U}{\partial Y} + \frac{\partial V}{\partial X} + \frac{\partial W}{\partial X} \frac{\partial W}{\partial Y} \end{Bmatrix} - Z \begin{Bmatrix} W_{,XX} \\ W_{,YY} \\ 2W_{,XY} \end{Bmatrix} \right) \quad (5)$$

where comma stands for the partial derivative, so the subscripts ' $i = (X, Y, XX, XY, YY)$ ' denote respectively  $(\frac{\partial}{\partial X}, \frac{\partial}{\partial Y}, \frac{\partial^2}{\partial X^2}, \frac{\partial^2}{\partial X \partial Y}, \frac{\partial^2}{\partial Y^2})$ . Inserting Eq. (4) into Eq. (A.4) leads the following relation between the deviatoric part of the couple stress and the displacements of the mid-plane.

$$\begin{Bmatrix} m_{XX} \\ m_{YY} \\ m_{XY} \\ m_{XZ} \\ m_{YZ} \end{Bmatrix} = \frac{El^2}{1+\nu} \begin{bmatrix} 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & \frac{1}{2} & 0 & 0 \\ 0 & 0 & 0 & \frac{1}{4} & 0 \\ 0 & 0 & 0 & 0 & \frac{1}{4} \end{bmatrix} \begin{Bmatrix} \frac{\partial^2 W}{\partial X \partial Y} \\ -\frac{\partial^2 W}{\partial X \partial Y} \\ \frac{\partial^2 W}{\partial Y^2} - \frac{\partial^2 W}{\partial X^2} \\ \frac{\partial^2 V}{\partial X^2} - \frac{\partial^2 U}{\partial X \partial Y} \\ \frac{\partial^2 V}{\partial X \partial Y} - \frac{\partial^2 U}{\partial Y^2} \end{Bmatrix} \quad (6)$$

The governing equations of motion and associated boundary conditions for the micro-plate can be derived by using Hamilton's principle (See Appendix B). Upon substitution of Eqs. (B.11a)–(B.11c) into Eqs. (B.9a)–(B.9c), the governing equations of motion can be obtained:

$$\begin{aligned} \frac{\partial^2 U}{\partial X^2} + \frac{1}{2}(1+\nu) \left( \frac{\partial^2 V}{\partial X \partial Y} + \frac{\partial W}{\partial Y} \frac{\partial^2 W}{\partial X \partial Y} \right) + \frac{1}{2}(1-\nu) \left( \frac{\partial^2 U}{\partial Y^2} + \frac{\partial W}{\partial X} \frac{\partial^2 W}{\partial Y^2} \right) \\ + \frac{\partial W}{\partial X} \frac{\partial^2 W}{\partial X^2} + \frac{l^2(1-\nu)}{8} \nabla^2 \left( \frac{\partial^2 V}{\partial X \partial Y} - \frac{\partial^2 U}{\partial Y^2} \right) = \frac{1-\nu^2}{Eh} l_0 \frac{\partial^2 U}{\partial t^2}, \end{aligned} \quad (7)$$

$$\begin{aligned} \frac{\partial^2 V}{\partial Y^2} + \frac{1}{2}(1+\nu) \left( \frac{\partial^2 U}{\partial X \partial Y} + \frac{\partial W}{\partial X} \frac{\partial^2 W}{\partial X \partial Y} \right) + \frac{1}{2}(1-\nu) \left( \frac{\partial^2 V}{\partial X^2} + \frac{\partial W}{\partial Y} \frac{\partial^2 W}{\partial X^2} \right) \\ + \frac{\partial W}{\partial Y} \frac{\partial^2 W}{\partial Y^2} + \frac{l^2(1-\nu)}{8} \nabla^2 \left( \frac{\partial^2 U}{\partial X \partial Y} - \frac{\partial^2 V}{\partial X^2} \right) = \frac{1-\nu^2}{Eh} I_0 \frac{\partial^2 V}{\partial t^2}, \end{aligned} \quad (7b)$$

$$\begin{aligned} \left( \mu h l^2 + \frac{E h^3}{12(1-\nu^2)} \right) \nabla^4 W + I_0 \frac{\partial^2 W}{\partial t^2} = I_2 \frac{\partial^2}{\partial t^2} (\nabla^2 W) + Y_{xx} \frac{\partial^2 W}{\partial X^2} \\ + 2Y_{xy} \frac{\partial^2 W}{\partial X \partial Y} + Y_{yy} \frac{\partial^2 W}{\partial Y^2} + N_{xx}^r \frac{\partial^2 W}{\partial X^2} + N_{yy}^r \frac{\partial^2 W}{\partial Y^2} + \frac{\varepsilon_0 V(t)^2}{2(g-W)^2} \\ - \frac{E h l^2}{8(1+\nu)} \left( \nabla^2 \left( \frac{\partial^2 V}{\partial X \partial Y} - \frac{\partial^2 U}{\partial Y^2} \right) \frac{\partial W}{\partial X} + \nabla^2 \left( \frac{\partial^2 U}{\partial X \partial Y} - \frac{\partial^2 V}{\partial X^2} \right) \frac{\partial W}{\partial Y} \right) \end{aligned} \quad (7c)$$

Eqs. (7a)–(7c) are the non-homogeneous form of the dynamic equations of micro-plate by the combined applied voltages AC and DC. Where the operators  $\nabla^2$  and  $\nabla^4$  in two-dimensional space can be stated as:

$$\nabla^2 = \left( \frac{\partial^2}{\partial X^2} + \frac{\partial^2}{\partial Y^2} \right), \quad (8)$$

$$\nabla^4 = \nabla^2 \nabla^2 = \left( \frac{\partial^4}{\partial X^4} + 2 \frac{\partial^4}{\partial X^2 \partial Y^2} + \frac{\partial^4}{\partial Y^4} \right) \quad (9)$$

Also, the MEM plates are often slender, i.e.,  $b > 100h$ , so the in-plane oscillations in comparison to the transverse vibration are quite small and insignificant. Also, it is easy to ignore the in-plane accelerations against the transverse plate acceleration [4,65]. Furthermore, in these structures, the transverse rotational acceleration can be neglected against its transmitted acceleration [65]. So, the inertia terms  $I_0 \frac{\partial^2 U}{\partial t^2}$  and  $I_0 \frac{\partial^2 V}{\partial t^2}$  in Eq. (7a) and (b) can be neglected. Furthermore, because of the slenderness of micro-plate, the rotary inertia term  $I_2 \frac{\partial^2}{\partial t^2} (\nabla^2 W)$  is also insignificant in comparison to the translatory one (i.e.,  $I_0 \frac{\partial^2 W}{\partial t^2}$ ) and can be ignored too [52]. It should be noted that the limitations of classical plate theory are depends on three different factors: the curvatures should be small, the in-plane plate dimensions should be large compared to the thickness and membrane strains can be neglected. So Eqs. (B.9a)–(B.9c) can be rewritten as follows:

$$\frac{\partial}{\partial X} Y_{xx} + \frac{\partial}{\partial Y} Y_{xy} + \frac{1}{2} \left( \frac{\partial^2}{\partial X \partial Y} \Gamma_{xz} + \frac{\partial^2}{\partial Y^2} \Gamma_{yz} \right) = 0, \quad (10)$$

$$\frac{\partial}{\partial X} Y_{xy} + \frac{\partial}{\partial Y} Y_{yy} - \frac{1}{2} \left( \frac{\partial^2}{\partial X^2} \Gamma_{xz} + \frac{\partial^2}{\partial X \partial Y} \Gamma_{yz} \right) = 0, \quad (10b)$$

$$\begin{aligned} \frac{\partial^2 \Xi_{xx}}{\partial X^2} + 2 \frac{\partial^2 \Xi_{xy}}{\partial X \partial Y} + \frac{\partial^2 \Xi_{yy}}{\partial Y^2} + Y_{xx} \frac{\partial^2 W}{\partial X^2} + 2Y_{xy} \frac{\partial^2 W}{\partial X \partial Y} + Y_{yy} \frac{\partial^2 W}{\partial Y^2} \\ + \frac{\partial Y_{xx}}{\partial X} \frac{\partial W}{\partial X} + \frac{\partial Y_{xy}}{\partial X} \frac{\partial W}{\partial Y} + \frac{\partial Y_{xy}}{\partial Y} \frac{\partial W}{\partial X} + \frac{\partial Y_{yy}}{\partial Y} \frac{\partial W}{\partial Y} + N_{xx}^r \frac{\partial^2 W}{\partial X^2} + N_{yy}^r \frac{\partial^2 W}{\partial Y^2} \\ + \frac{\partial^2}{\partial X^2} \Gamma_{xy} - \frac{\partial^2}{\partial Y^2} \Gamma_{xy} + \frac{\partial^2}{\partial X \partial Y} \Gamma_{yy} - \frac{\partial^2}{\partial X \partial Y} \Gamma_{xx} + \frac{\varepsilon_0 V(t)^2}{2(g-W)^2} = I_0 \ddot{W} \end{aligned} \quad (10c)$$

By introducing the stress functions as follow, Eq. (10a) and (b) are automatically satisfied.

$$Y_{xx} = \phi_{,yy}, Y_{yy} = \phi_{,xx}, Y_{xy} = -\phi_{,xy}, \Gamma_{xz} = \phi_{,y}, \Gamma_{yz} = -\phi_{,x} \quad (11)$$

So, Eq. (10c) is converted as follows:

$$\begin{aligned} \frac{\partial^2}{\partial X^2} \Xi_{xx} + 2 \frac{\partial^2}{\partial X \partial Y} \Xi_{xy} + \frac{\partial^2}{\partial Y^2} \Xi_{yy} + \phi_{,yy} \frac{\partial^2 W}{\partial X^2} - 2\phi_{,xy} \frac{\partial^2 W}{\partial X \partial Y} + \phi_{,xx} \frac{\partial^2 W}{\partial Y^2} \\ + \frac{\partial(\phi_{,yy})}{\partial X} \frac{\partial W}{\partial X} - \frac{\partial(\phi_{,xy})}{\partial X} \frac{\partial W}{\partial Y} - \frac{\partial(\phi_{,xy})}{\partial Y} \frac{\partial W}{\partial X} + \frac{\partial(\phi_{,xx})}{\partial Y} \frac{\partial W}{\partial Y} + N_{xx}^r \frac{\partial^2 W}{\partial X^2} \\ + N_{yy}^r \frac{\partial^2 W}{\partial Y^2} + \frac{\partial^2}{\partial X^2} \Gamma_{xy} - \frac{\partial^2}{\partial Y^2} \Gamma_{xy} + \frac{\partial^2 \Gamma_{yy}}{\partial X \partial Y} - \frac{\partial^2 \Gamma_{xx}}{\partial X \partial Y} + \frac{\varepsilon_0 V(t)^2}{2(g-W)^2} = I_0 \ddot{W} \end{aligned} \quad (12)$$



By replacing  $\Xi_i, \Gamma_i$   $i = XX, YY, XY$  from the relations (B.11a) and (B.11b), the following equation is obtained:

$$\begin{aligned} & \frac{\partial^2}{\partial X^2} \left( \frac{-Eh^3}{12(1-\nu^2)} \frac{\partial^2 W}{\partial X^2} + \frac{-Eh^3 \nu}{12(1-\nu^2)} \frac{\partial^2 W}{\partial Y^2} \right) + 2 \frac{\partial^2}{\partial X \partial Y} \left( \frac{-Eh^3}{12(1+\nu)} \frac{\partial^2 W}{\partial Y \partial X} \right) \\ & + \frac{\partial^2}{\partial Y^2} \left( \frac{-Eh^3}{12(1-\nu^2)} \frac{\partial^2 W}{\partial Y^2} + \frac{-Eh^3 \nu}{12(1-\nu^2)} \frac{\partial^2 W}{\partial X^2} \right) + \phi_{,YY} \frac{\partial^2}{\partial X^2} W - 2\phi_{,XY} \frac{\partial^2 W}{\partial X \partial Y} \\ & + \phi_{,XX} \frac{\partial^2}{\partial Y^2} W + \frac{\partial(\phi_{,YY})}{\partial X} \frac{\partial W}{\partial X} - \frac{\partial(\phi_{,XY})}{\partial X} \frac{\partial W}{\partial Y} - \frac{\partial(\phi_{,XY})}{\partial Y} \frac{\partial W}{\partial X} + \frac{\partial(\phi_{,XX})}{\partial Y} \frac{\partial W}{\partial Y} \\ & + N_{XX}^r \frac{\partial^2 W}{\partial X^2} + N_{YY}^r \frac{\partial^2 W}{\partial Y^2} - \frac{1}{2} \frac{Ehl^2}{1+\nu} \frac{\partial^2}{\partial X^2} \left( \frac{\partial^2 W}{\partial X^2} - \frac{\partial^2 W}{\partial Y^2} \right) \\ & + \frac{1}{2} \frac{Ehl^2}{1+\nu} \frac{\partial^2}{\partial Y^2} \left( \frac{\partial^2 W}{\partial X^2} - \frac{\partial^2 W}{\partial Y^2} \right) - \frac{Ehl^2}{1+\nu} \frac{\partial^2}{\partial X \partial Y} \left( \frac{\partial^2 W}{\partial X \partial Y} \right) \\ & - \frac{Ehl^2}{1+\nu} \frac{\partial^2}{\partial X \partial Y} \left( \frac{\partial^2 W}{\partial X \partial Y} \right) + \frac{\varepsilon_0 V(t)^2}{2(g-W)^2} = I_0 \ddot{W} \end{aligned} \quad (13)$$

By removing  $U$  and  $V$  from Eq. (2), the compatibility equation is obtained as follows:

$$\frac{\partial^2 \varepsilon_{XX}^{(0)}}{\partial Y^2} + \frac{\partial^2 \varepsilon_{YY}^{(0)}}{\partial X^2} - \frac{\partial^2 \varepsilon_{XY}^{(0)}}{\partial X \partial Y} = \left( \frac{\partial^2 W}{\partial X \partial Y} \right)^2 - \frac{\partial^2 W}{\partial X^2} \frac{\partial^2 W}{\partial Y^2} \quad (14)$$

By solving Eqs. (B.11a) and (B.11c) for  $\varepsilon_k^{(0)}$ ,  $k = XX, YY, XY$  and replacing Eq. (11), the following relations are obtained:

$$\varepsilon_{XX}^{(0)} = \frac{1}{Eh} (\phi_{,YY} - \nu \phi_{,XX}), \quad (15)$$

$$\varepsilon_{YY}^{(0)} = \frac{1}{Eh} (-\nu \phi_{,YY} + \phi_{,XX}), \quad (15b)$$

$$\varepsilon_{XY}^{(0)} = -2 \frac{1+\nu}{Eh} \phi_{,XY} \quad (15c)$$

Replacing the relationships (15a)–(c) in Eq. (14) and assume that  $\phi$  is replaced by  $\phi = hF$ , in which  $F$  is the Airy stress function, the following equations are obtained:

$$\nabla^4 F = E(W_{,XY}^2 - W_{,XX}W_{,YY}) \quad (16)$$

By simplifying Eq. (13), Eq. (17) is obtained as follows:

$$\begin{aligned} \Lambda(W, F) & \equiv h \left( F_{,YY} \frac{\partial^2}{\partial X^2} W - 2F_{,XY} \frac{\partial^2}{\partial X \partial Y} W + F_{,XX} \frac{\partial^2}{\partial Y^2} W \right) + N_{XX}^r \frac{\partial^2 W}{\partial X^2} \\ & + N_{YY}^r \frac{\partial^2 W}{\partial Y^2} + q - I_0 \ddot{W} - (\nabla^4 W) D_{eq} = 0 \quad \& D_{eq} = \left( \frac{1}{2} \frac{Ehl^2}{1+\nu} + \frac{Eh^3}{12(1-\nu^2)} \right), \\ q(X, Y) & = \frac{\varepsilon_0 V(t)^2}{2(g-W(X, Y))^2} \end{aligned} \quad (17)$$

Eqs. (16) and (17) are governing equations of micro-plate regarding  $W$  and  $F$ .

The associated out-of-plane boundary conditions for clamped micro-plate with immovable edges have the form as:

$$\begin{aligned} W = W_{,X} = 0 \quad \text{at} \quad X = \pm \frac{a}{2} \\ W = W_{,Y} = 0 \quad \text{at} \quad Y = \pm \frac{b}{2} \end{aligned} \quad (18)$$

The in-plane conditions for clamped boundary condition are considered as (All edges immovably constrained):

$$\begin{aligned} U = F_{,XY} = 0 \quad \text{at} \quad X = \pm \frac{a}{2} \\ V = F_{,XY} = 0 \quad \text{at} \quad Y = \pm \frac{b}{2} \end{aligned} \quad (18b)$$

Since the effect of shear deformation on the resonance frequencies of thin micro plates is not negligible, one may suspect that there are other thin-plate geometries where in-plane motion is important. So, in this condition resonant frequencies

predicted by classical plate theory would be agree well with those measured by first shear deformation theory or higher plate theories. According to what is commonly stated in the literature, when in-plane motion is restricted one can assume that classical plate theory is applicable simply because the plate is thin [66].

Eqs. (16) and (17) should be solved in conjunction with boundary conditions (18a, 18b).

### 3. Solution procedure

Approximate methods for rectangular and circular plates with different boundary conditions were obtained by Yamaki [67]. Kung and Pao [68] used the combination of the Galerkin method and the Harmonic Balance Method (HBM) to analyze the vibration of buckled rectangular plates. Using the method of multiple scales, Hadian and Nayfeh [69] studied the response of circular plates under intermittent external stimulation. Shi and Mei [70] studied large amplitude free vibration of plates using the Reduced Order method (ROM). A large number of researchers have used the combination of finite element method and Harmonic Balance method to study the nonlinear geometric vibration of thin isotropic plates. The dynamical behavior of the plates for large-amplitude vibrations by theoretical and laboratory methods has been investigated by Benamar et al. [71]. For further study on the nonlinear behavior of plates, one can refer to nonlinear vibration and stability books of shells and plates [72], nonlinear vibrations, nonlinear analysis of plates [73], and linear and nonlinear mechanical mechanisms [30].

#### 3.1. Static case

Because of the limitation in applying an electrical voltage to the system, the study of static behavior of micro-plate is essential to identify the maximum DC voltage and prevent the static instability of the system. For studying static behavior, a solution is assumed in the form of a generalized double Fourier series for the static case, i.e.,  $I_0 \dot{W} = 0$  [74].

$$W = \sum_{m=1}^{\infty} \sum_{n=1}^{\infty} W_{mn} x_m(X) y_n(Y), \quad F = \sum_{p=1}^{\infty} \sum_{q=1}^{\infty} F_{pq} x_p(X) y_q(Y) \quad (19)$$

where  $W_{mn}$  and  $F_{pq}$  are constant coefficients to be determined and  $x_m$ ,  $x_p$  are beam Eigen functions given by:

$$x_m = \frac{\cosh \alpha_m X}{\cosh \alpha_m \frac{a}{2}} - \frac{\cos \alpha_m X}{\cos \alpha_m \frac{a}{2}}, \quad y_n = \frac{\cosh \beta_n Y}{\cosh \beta_n \frac{b}{2}} - \frac{\cos \beta_n Y}{\cos \beta_n \frac{b}{2}} \quad (20)$$

All the boundary conditions (18) are satisfied if the values of  $\alpha_m$  and  $\beta_n$  be the roots of the transcendental equation:

$$\tanh \lambda_m + \tan \lambda_m = 0 \quad (21)$$

Where

$$\lambda_m = \alpha_m \frac{a}{2} \quad \text{or} \quad \beta_n \frac{b}{2} \quad (22)$$

The roots of Eq. (21) are obtained simplicity. The functions  $x_m(X)$  and  $y_n(Y)$  satisfy the following orthogonality relations:

$$\int_{-\frac{a}{2}}^{\frac{a}{2}} x_i x_j dX = \begin{cases} 0 & i \neq j \\ a & i = j \end{cases}, \quad \int_{-\frac{b}{2}}^{\frac{b}{2}} y_i y_j dY = \begin{cases} 0 & i \neq j \\ b & i = j \end{cases} \quad (23)$$

The transverse load  $q(X, Y)$  is expanded into a double series:

$$q(X, Y) = \sum_{m=1}^{\infty} \sum_{n=1}^{\infty} q_{mn} x_m(X) y_n(Y) \quad (24)$$

Where

$$q_{mn} = \frac{1}{ab} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} q(X, Y) x_m(X) y_n(Y) dX dY \quad (25)$$

Substituting Eqs. (19) and (24) into Eqs. (17) and (16) leads to:

$$\begin{aligned} \sum_{m=1}^{\infty} \sum_{n=1}^{\infty} W_{mn} ((\alpha_m^4 + \beta_n^4) x_m y_n + 2x''_m y''_n) &= \frac{1}{D_{eq}} \\ &\left( \sum_{m=1}^{\infty} \sum_{n=1}^{\infty} q_{mn} x_m y_n + h \sum_{p=1}^{\infty} \sum_{q=1}^{\infty} \sum_{r=1}^{\infty} \sum_{s=1}^{\infty} W_{pqrs} (x_r x''_p y_q y''_s + x_p x''_r y_s y''_q - 2x'_p x'_r y'_q y'_s) \right. \\ &\quad \left. + N_{XX}^r \sum_{m=1}^{\infty} \sum_{n=1}^{\infty} W_{mn} \alpha_m^2 x_m y_n + N_{YY}^r \sum_{m=1}^{\infty} \sum_{n=1}^{\infty} W_{mn} \beta_n^2 x_m y_n \right) \end{aligned} \quad (26)$$



186

$$\begin{aligned} & \sum_{p=1}^{\infty} \sum_{q=1}^{\infty} F_{pq} \left( (\alpha_p^4 + \beta_q^4) x_p y_q + 2x''_p y''_q \right) \\ &= E \sum_{m=1}^{\infty} \sum_{n=1}^{\infty} \sum_{r=1}^{\infty} \sum_{s=1}^{\infty} W_{mn} W_{rs} \left( x'_m x'_r y'_n y'_s - x_r x''_m y_n y''_s \right) \end{aligned} \quad (27)$$

187 in which primes denote differential with respect to the corresponding coordinates. Multiplying each of Eqs. (26) and (27) by  
 188  $x_i(X) \times y_j(Y)$ , integrating with respect to  $X$  and  $Y$  over their respective intervals, and using Eqs. (22) and (23) leads to a  
 189 system of nonlinear algebraic equations.

$$\begin{aligned} & W_{ij} \left( \frac{b^2}{a^2} \lambda_i^4 + \frac{a^2}{b^2} \lambda_j^4 \right) + 2 \sum_{m=1}^{\infty} \sum_{n=1}^{\infty} W_{mn} \lambda_m^2 \lambda_n^2 K_1^{im} K_1^{jn} = \frac{a^2 b^2 q_{ij}}{16 D_{eq}} + \frac{h}{D_{eq}} \\ & \times \sum_{p=1}^{\infty} \sum_{q=1}^{\infty} \sum_{r=1}^{\infty} \sum_{s=1}^{\infty} W_{pq} F_{rs} \cdot \left( \lambda_p^2 \lambda_s^2 K_2^{irp} L_2^{jqs} + \lambda_q^2 \lambda_r^2 K_2^{ipr} L_2^{jsq} - 2 \lambda_p \lambda_q \lambda_r \lambda_s K_3^{ipr} L_3^{jqs} \right), \\ & i, j = 1, 2, 3, \dots \end{aligned} \quad (28)$$

190

$$\begin{aligned} & F_{ij} \left( \frac{b^2}{a^2} \lambda_i^4 + \frac{a^2}{b^2} \lambda_j^4 \right) + 2 \sum_{p=1}^{\infty} \sum_{q=1}^{\infty} F_{pq} \lambda_p^2 \lambda_q^2 K_1^{ip} K_1^{jq} \\ &= E \sum_{m=1}^{\infty} \sum_{n=1}^{\infty} \sum_{r=1}^{\infty} \sum_{s=1}^{\infty} W_{mn} W_{rs} \cdot \left( \lambda_m \lambda_n \lambda_r \lambda_s K_3^{imr} L_3^{jns} - \lambda_m^2 \lambda_s^2 K_2^{irm} L_2^{jns} \right), \quad i, j = 1, 2, 3, \dots \end{aligned} \quad (29)$$

191 where

$$\begin{aligned} & K_1^{im} = \frac{1}{a \alpha_m^2} \int_{-\frac{a}{2}}^{\frac{a}{2}} x_i x''_m dX, \quad L_1^{jn} = \frac{1}{b \beta_n^2} \int_{-\frac{b}{2}}^{\frac{b}{2}} y_j y''_n dY, \quad K_2^{irp} = \frac{1}{a \alpha_p^2} \int_{-\frac{a}{2}}^{\frac{a}{2}} x_i x_r x''_p dX, \\ & L_2^{jqs} = \frac{1}{b \beta_s^2} \int_{-\frac{b}{2}}^{\frac{b}{2}} y_j y_q y''_s dY, \quad K_3^{ipr} = \frac{1}{a \alpha_p \alpha_r} \int_{-\frac{a}{2}}^{\frac{a}{2}} x_i x'_p x'_r dX, \quad L_3^{jqs} = \frac{1}{b \beta_q \beta_s} \int_{-\frac{b}{2}}^{\frac{b}{2}} y_j y'_q y'_s dY \end{aligned} \quad (30)$$

192 Let  $q(X, Y)$  be load  $P$  uniformly distributed over a portion of the plate surface. The load coefficients introduced in Eq. (25) be-  
 193 come:

$$q_{mn} = \frac{P}{\alpha \beta \lambda_m \lambda_n} (\sinh \lambda_m - \sin \lambda_m) (\sinh \lambda_n - \sin \lambda_n) \quad (31)$$

194 By substituting Eq. (31) into Eq. (28), the deflection vector and stress resultants can be calculated. With these values of  $q_{mn}$ ,  
 195 the geometrically nonlinear behavior of a rectangular micro-plate under a uniformly distributed load ( $P$ ) can be investigated  
 196 by the previous series solution. Similarly, the solution can be applied to other types of transverse loading.

### 197 3.1.1. Galerkin's method for clamped micro-plate

198 The large deflection of a rectangular clamped micro-plate under distributed load  $q(X, Y)$  is reconsidered by making use  
 199 of the one-term approximation of the Galerkin method [75]. The equilibrium Eq. (17) and compatibility relation (16) and  
 200 boundary conditions (38) remain unchanged. The transverse deflection is assumed to be of the form Yeh and Liu [76]

$$W = g W_m \cos^2 \left( \frac{\pi X}{a} \right) \cos^2 \left( \frac{\pi Y}{b} \right) \quad (32)$$

201 in which  $W_m$ , is the non-dimensional maximum deflection at the plate center given by  $W_{0/g}$ , and  $W_0$  denotes the central de-  
 202 flection. This approximated deflection obviously satisfies the geometrical boundary conditions in Eq. (18). Upon substitution  
 203 Eq. (16) may be expressed as:

$$\nabla^4 F = - \frac{\pi^4 E g^2 W_m^2}{a^2 b^2} \sum_{p=0}^{\infty} \sum_{q=0}^{\infty} a_{pq} R_p(X) S_q(Y) \quad (33)$$

204 Where

$$\begin{aligned} & R_p(X) = \cos \frac{2p\pi X}{a}, \quad S_q(Y) = \cos \frac{2q\pi Y}{b}, \quad a_{01} = a_{10} = a_{02} = a_{20} = a_{12} = \\ & a_{21} = \frac{1}{2}, \quad a_{11} = 1 \quad \text{and all other } a_{pq} = 0 \end{aligned} \quad (34)$$

205 The general solution of Eq. (33) is the sum of the complementary function  $F_c$  and a particular integral  $F_p$ , i.e.,  $F = F_c + F_p$ .  
 206 A particular solution of Eq. (33) may be expressed as:

$$F_p = Eg^2 W_m^2 \sum_{p=0}^{\infty} \sum_{q=0}^{\infty} b_{pq} R_p(X) S_q(Y) \quad \& b_{pq} = -\frac{\psi^2 a_{pq}}{16(p^2 + \psi^2 q^2)^2}, \quad \psi = \frac{a}{b} \quad (35)$$

207 It is easily seen that  $F_p$  given by expression (35) is an even function of  $X$  and  $Y$  with the vanishing shear stresses along the  
 208 boundaries. With the same properties, the complementary function may be expressed in the form:

$$F_c = \frac{W_m^2}{2} (C_1 X^2 + C_2 Y^2) + Eg^2 W_m^2 \sum_{n=1}^{\infty} \left\{ \frac{A_n}{n^2 [\sinh(n\pi/\psi) \cosh(n\pi/\psi) + n\pi/\psi]} \right. \\ \left[ \left( \sinh \frac{n\pi}{\psi} + \frac{n\pi}{\psi} \cosh \frac{n\pi}{\psi} \right) \cosh \frac{2n\pi}{a} Y - \frac{2n\pi}{a} Y \sinh \frac{n\pi}{\psi} \sinh \frac{2n\pi}{a} Y \right] \cos \frac{2n\pi X}{a} \\ + \frac{B_n}{n^2 \psi^2 [\sinh n\pi \psi \cosh n\pi \psi + n\pi \psi]} [(\sinh n\pi \psi + n\pi \psi \cosh n\pi \psi) \cosh \frac{2n\pi X}{b} \\ - \frac{2n\pi}{b} X \sinh n\pi \psi \sinh \frac{2n\pi X}{b}] \cos \frac{2n\pi Y}{b} \Big\} \quad (36)$$

209 In which  $C_1$ ,  $C_2$ ,  $A_n$ , and  $B_n$ , are arbitrary constants. (See Appendix C, for more information). For convenience, the comple-  
 210 mentary function  $F_c$  is also represented by a double cosine series.

$$F_c = Eg^2 W_m^2 \sum_{p=0}^{\infty} \sum_{q=0}^{\infty} c_{pq} R_p(X) S_q(Y) \quad (37)$$

211 The Fourier coefficients  $c_{mn}$  in the series are given by:

$$c_{mn} = \frac{4\psi}{\pi (m^2 + \psi^2 n^2)^2} \\ \times \left( \frac{m(-1)^n \zeta_n \sinh^2(m\pi/\psi) A_m}{(m\pi/\psi) + \sinh(m\pi/\psi) \cosh(m\pi/\psi)} + \frac{n(-1)^m \zeta_m \sinh^2(n\pi\psi B_n)}{n\pi\psi + \sinh n\pi\psi \cosh n\pi\psi} \right) \\ m, n = 0, 1, 2, \dots \quad (38)$$

212 where

$$\zeta_0 = \frac{1}{2}, \quad \zeta_1 = \zeta_2 = \dots = 1 \quad (39)$$

213 By substitution ( $F = F_c + F_p$ ) yields:

$$F = Eg^2 W_m^2 \sum_{p=0}^{\infty} \sum_{q=0}^{\infty} (b_{pq} + c_{pq}) R_p(X) S_q(Y) \quad (40)$$

214 Function (32) for  $W$  and Function (36) for  $F$  satisfy all the boundary conditions (18) as well as the compatibility condition  
 215 (16). With these expressions, however, the equilibrium Eq. (16) generally cannot be precisely satisfied. Instead of satisfaction  
 216 of this equation, we apply the Galerkin method to minimize the error function which is obtained by inserting Eqs. (32) and  
 217 (40) into (16), i.e.,

$$\iint_A \Delta(W, F) W \, dX \, dY = 0 \quad (41)$$

### 218 3.2. Vibrational of micro-plate due to the harmonic electrical force

219 The in-plane displacements  $U$  and  $V$  are related to the stress function ( $F$ ) by the following equation:

$$U = \int_0^X \left( \frac{(F_{YY} - \nu F_{XX})}{E} - \frac{1}{2} (W_{,X})^2 \right) dX \quad \& \quad V = \int_0^Y \left( \frac{(F_{XX} - \nu F_{YY})}{E} - \frac{1}{2} (W_{,Y})^2 \right) dY \quad (42)$$

220 A one-term approximate solution of the governing Eqs. (16) and (17) satisfying the prescribed boundary conditions in  
 221 each case which formulated by application of the Galerkin method. The conditions (18a) are satisfied by assuming the  
 222 deflection function as [77]:

$$W = gR(t) \cos^2\left(\frac{\pi X}{a}\right) \cos^2\left(\frac{\pi Y}{b}\right) \quad (43)$$

223 in which  $R(t)$  is a function of time ( $t$ ) with its maximum value being:

$$R_{\max}(t) = \frac{W_m}{g} \quad (44)$$

In this expression  $W_m$ , is the maximum deflection of the plate. Substituting expressions (43) into the compatibility Eq. (16) yields:

$$\nabla^4 F = \frac{\pi^4 E g^2 R^2}{a^2 b^2} \sum_{p=0}^{\infty} \sum_{q=0}^{\infty} a_{pq} \cos \frac{2p\pi X}{a} \cos \frac{2q\pi Y}{b} \quad (45)$$

in which  $a_{pq}$  are known Fourier coefficients. The general solution of Eq. (45) is  $F = F_c + F_p$ , where  $F_c$  is the complementary function and  $F_p$  is a particular solution which may be expressed as:

$$F_p = E g^2 R^2 \sum_{p=0}^{\infty} \sum_{q=0}^{\infty} b_{pq} \cos \frac{2p\pi X}{a} \cos \frac{2q\pi Y}{b}, \quad b_{pq} = \frac{\psi^2 a_{pq}}{16(p^2 + \psi^2 q^2)^2} \quad (46)$$

After some manipulation the nonzero coefficients  $b_{pq}$  are obtained:

$$\begin{aligned} b_{01} &= -\frac{1}{32\psi^2}, \quad b_{10} = -\frac{\psi^2}{32}, \quad b_{11} = -\frac{\psi^2}{16(1+\psi^2)^2}, \quad b_{02} = -\frac{1}{512\psi^2}, \\ b_{20} &= -\frac{\psi^2}{512}, \quad b_{12} = -\frac{\psi^2}{32(1+4\psi^2)^2}, \quad b_{21} = -\frac{\psi^2}{32(4+\psi^2)^2} \end{aligned} \quad (47)$$

It is observed that  $F_p$ , is an even function in  $X$  and  $Y$  satisfying the condition for zero shear stress along the edges of the plate. The constants  $C_1$ ,  $C_2$ ,  $A_n$  and  $B_n$  in Eq. (36) are to be determined by the in-plane boundary conditions. Inserting the expressions  $W$  and  $F$  in conditions (18b), by use of Eq. (42), for an immovable micro-plate, these constants obtained as:

$$C_1 = \frac{3}{32} \frac{\pi^2 E g^2 (\psi^2 + \nu)}{a^2 (1 - \nu^2)}, \quad C_2 = \frac{3}{32} \frac{\pi^2 E g^2 (\nu \psi^2 + 1)}{a^2 (1 - \nu^2)}, \quad A_n = B_n = 0 \quad (48)$$

Now the function  $F$  can be written in the general form:

$$F = \frac{R^2}{2} (C_1 X^2 + C_2 Y^2) + E g^2 R^2 \sum_{p=0}^{\infty} \sum_{q=0}^{\infty} b_{pq} \cos \frac{2p\pi X}{a} \cos \frac{2q\pi Y}{b} \quad (49)$$

Finally, the following equation is obtained for  $F$ :

$$\begin{aligned} F &= \frac{R^2}{2} \left( \frac{3}{32} \frac{\pi^2 E g^2 (\psi^2 + \nu)}{a^2 (1 - \nu^2)} X^2 + \frac{3}{32} \frac{\pi^2 E g^2 (\nu \psi^2 + 1)}{a^2 (1 - \nu^2)} Y^2 \right) - E g^2 R^2 \\ &\times \left( \frac{1}{32\psi^2} \cos \frac{2\pi Y}{b} + \frac{\psi^2}{32} \cos \frac{2\pi X}{a} + \frac{\psi^2}{16(1+\psi^2)^2} \cos \frac{2\pi X}{a} \cos \frac{2\pi Y}{b} \right. \\ &+ \frac{1}{512\psi^2} \cos \frac{4\pi Y}{b} + \frac{\psi^2}{512} \cos \frac{4\pi X}{a} + \frac{\psi^2}{32(1+4\psi^2)^2} \cos \frac{2\pi X}{a} \cos \frac{4\pi Y}{b} \\ &\left. + \frac{\psi^2}{32(4+\psi^2)^2} \cos \frac{4\pi X}{a} \cos \frac{2\pi Y}{b} \right) \end{aligned} \quad (50)$$

### 3.3. Non-dimensionalization of the governing equations

It is an excellent practice to non-dimensionalize the governing equations before treating them with perturbation methods to simplify and avoid calculation errors. To this end, the non-dimensionalization parameters such as characteristic length, time and other non-dimensional variables are as follow:

$$\begin{aligned} x &= \frac{X}{a}, \quad y = \frac{Y}{b}, \quad w = \frac{W}{g}, \quad \psi = \frac{a}{b}, \quad \xi = 6(1 - \nu) \left( \frac{l}{h} \right)^2, \quad f = \frac{F}{E g^2}, \quad \kappa = \frac{g}{h}, \\ N_{i=x,y} &= \frac{12a^2(1 - \nu^2)N_i^r}{E h^3}, \quad \hat{t} = \frac{t}{a^2} \sqrt{\frac{E h^3}{12(1 - \nu^2)\rho h}}, \quad \beta = \frac{6(1 - \nu^2)\varepsilon_0 a^4}{E h^3 g^3} \end{aligned} \quad (51)$$

Upon substitution of the dimensionless quantities given in Eq. (51) into Eqs. (16), (17) and (50) moreover, multiplication both side Eqs. (16) and (17) by  $\frac{a^4}{g} \frac{12(1 - \nu^2)}{E h^3}$  and Eq. (50) by  $\frac{a^4}{E g^2}$  the following equations will be gained:

$$12\psi^2 \kappa^2 (1 - \nu^2) \left( f_{,yy} \frac{\partial^2}{\partial x^2} w - 2f_{,xy} \frac{\partial^2}{\partial x \partial y} w + f_{,xx} \frac{\partial^2}{\partial y^2} w \right) + N_x \frac{\partial^2 w}{\partial x^2}$$

$$\begin{aligned}
& + N_y \psi^2 \frac{\partial^2 w}{\partial y^2} + \frac{\beta V(t)^2}{(1-w)^2} - \frac{\partial^2 w}{\partial t^2} = \frac{\partial^4 w}{\partial x^4} + 2\psi^2 \frac{\partial^4 w}{\partial x^2 \partial y^2} + \psi^4 \frac{\partial^4 w}{\partial y^4} \\
& + \xi \left( \frac{\partial^4 w}{\partial x^4} + 2\psi^2 \frac{\partial^4 w}{\partial x^2 \partial y^2} + \psi^4 \frac{\partial^4 w}{\partial y^4} \right), \quad (52)
\end{aligned}$$

$$\frac{\partial^4 f}{\partial x^4} + 2\psi^2 \frac{\partial^4 f}{\partial x^2 \partial y^2} + \psi^4 \frac{\partial^4 f}{\partial y^4} + \psi^2 \left( \frac{\partial^2 w}{\partial x^2} \frac{\partial^2 w}{\partial y^2} - \left( \frac{\partial^2 w}{\partial x \partial y} \right)^2 \right) = 0, \quad (53)$$

$$\begin{aligned}
f = & \frac{R^2 \pi^2}{2} \left( \frac{3}{32} \frac{(\psi^2 + \nu)}{(1 - \nu^2)} x^2 + \frac{3}{32} \frac{(\nu \psi^2 + 1)}{(1 - \nu^2)} y^2 \right) - R^2 \\
& \times \left( \frac{1}{32\psi^2} \cos 2\pi y + \frac{\psi^2}{32} \cos 2\pi x + \frac{\psi^2}{16(1 + \psi^2)^2} \cos 2\pi x \cos 2\pi y \right. \\
& + \frac{1}{512\psi^2} \cos 4\pi y + \frac{\psi^2}{512} \cos 4\pi x + \frac{\psi^2}{32(1 + 4\psi^2)^2} \cos 2\pi x \cos 4\pi y \\
& \left. + \frac{\psi^2}{32(4 + \psi^2)^2} \cos 4\pi x \cos 2\pi y \right) \quad (54)
\end{aligned}$$

Eqs. (52) and (53) are equations of motion for the rectangular micro-plates under electric force in terms of  $w$  and  $f$ . In other words, Eq. (52) states the movement of the micro-plate in the  $Z$  direction and Eq. (53) is the plate compatibility equation. It can be understood that the length of the plate has a substantial effect on the static and dynamic components of the electrical force  $\beta$  (see Eq. (51)).

### 3.3.1. Non-dimensionalization of the boundary conditions

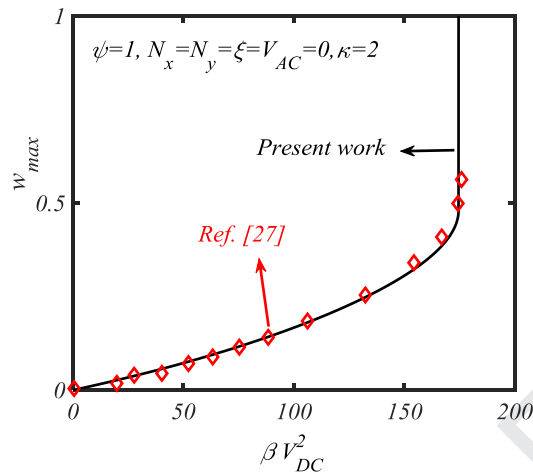
The associated in-plane and out-plane dimensionless boundary conditions for micro-plate with immovable edges have the form as [52, 54]

$$\begin{aligned}
\delta u &= 0 \quad \text{at } x = -\frac{1}{2}, \frac{1}{2} \quad \& \quad y = -\frac{1}{2}, \frac{1}{2} \\
\frac{\partial \delta u}{\partial x} &= 0 \quad \text{at } y = -\frac{1}{2}, \frac{1}{2} \\
\frac{\partial \delta u}{\partial y} &= 0 \quad \text{at } x = -\frac{1}{2}, \frac{1}{2} \quad \& \quad y = -\frac{1}{2}, \frac{1}{2} \\
\delta v &= 0 \quad \text{at } x = -\frac{1}{2}, \frac{1}{2} \quad \& \quad y = -\frac{1}{2}, \frac{1}{2} \\
\frac{\partial \delta v}{\partial x} &= 0 \quad \text{at } x = -\frac{1}{2}, \frac{1}{2} \quad \& \quad y = -\frac{1}{2}, \frac{1}{2} \\
\frac{\partial \delta v}{\partial y} &= 0 \quad \text{at } x = -\frac{1}{2}, \frac{1}{2} \\
\delta w &= 0 \quad \text{at } x = -\frac{1}{2}, \frac{1}{2} \quad \& \quad y = -\frac{1}{2}, \frac{1}{2} \\
\frac{\partial \delta w}{\partial x} &= 0 \quad \text{at } x = -\frac{1}{2}, \frac{1}{2} \quad \& \quad y = -\frac{1}{2}, \frac{1}{2} \\
\frac{\partial \delta w}{\partial y} &= 0 \quad \text{at } x = -\frac{1}{2}, \frac{1}{2} \quad \& \quad y = -\frac{1}{2}, \frac{1}{2} \quad (55)
\end{aligned}$$

## 4. Obtaining a set of ordinary differential equation

The large static deflection of the plate can be treated as a special case of the nonlinear plate vibration. In the discretization methods, one postulates the solution in the form  $w(x, y, t) = \sum_{m=1}^M \phi_m(x, y) R_m(t)$ , where  $M$  is a finite integer and  $\phi_m$  are the generalized coordinates [62]. By replacing  $w = R(t) \cos^2(\pi x) \cos^2(\pi y)$  and  $f$  from Eq. (54), into Eq. (52), multiplying the resulting equation in  $\cos^2(\pi x) \cos^2(\pi y)$  and integrating from  $(x, y) = (-\frac{1}{2}, -\frac{1}{2})$  to  $(x, y) = (\frac{1}{2}, \frac{1}{2})$ , the following non-linear ordinary differential equation is obtained:

$$\ddot{R}(t) + A_1 R(t) \ddot{R}(t) + A_2 R(t)^2 \ddot{R}(t) + A_3 R(t) + A_4 R(t)^2 + A_5 R(t)^3 + A_6 R(t)^4$$



**Fig. 2.** Comparison between the static deflection obtained by the current method for a micro-plate.

$$+A_7R(t)^5 + A_8 = 0 \quad (56)$$

The coefficients  $A_i$ ,  $i = 1, 2, \dots, 8$  are presented in Appendix D.

## 5. Result and discussion

The results presented in the next sections have been obtained for clamped micro-plates without in-plate motion and with the following coefficients  $N_x = N_y = 1$ ,  $\nu = 0.33$ , wherever these coefficients have not been specified.

### 5.1. Validation

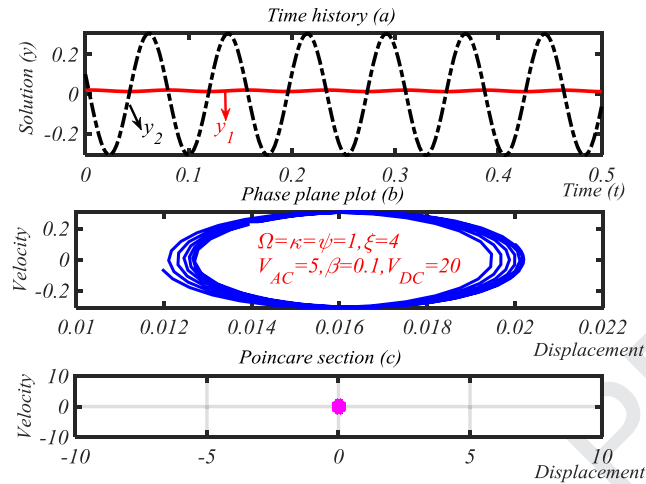
To measure the accuracy of the model proposed in this research, the following results are validated with the information contained in the previously published article. Fig. 2 shows a comparison between the static deflection obtained by the current method and those obtained by Zhao using a reduced order model based on analytically obtained basis functions [29]. As it is evident, there is a good agreement.

### 5.2. Time response and phase portrait

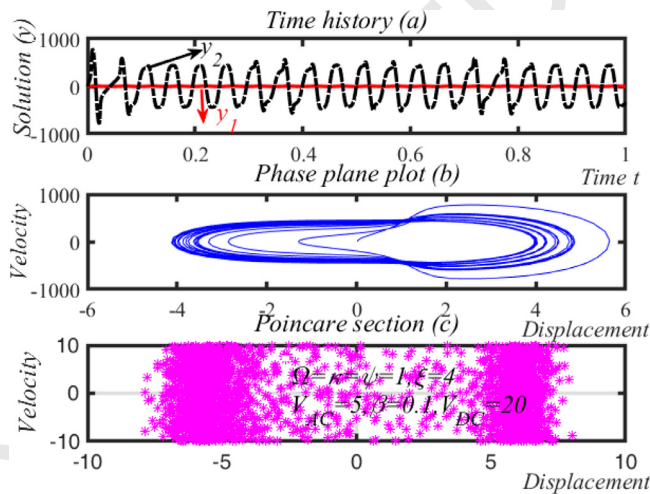
According to the Eq. (56) and by using a change of variables, i.e.,  $R(t) = y_1$ ,  $\dot{R}(t) = y_2$ , the following equations are obtained as:

$$\begin{aligned} \dot{y}_2 &= \frac{1}{(1 + A_1 y_1 + A_2 y_1^2)} \left( -(A_3 y_1 + A_4 y_1^2 + A_5 y_1^3 + A_6 y_1^4 + A_7 y_1^5) + \right. \\ &\quad \left. \frac{16\beta}{9} \left( V_{DC}^2 + 2V_{DC}V_{AC} \cos(\omega_1 t) + \frac{V_{AC}^2}{2} (1 + \cos(\omega_2 t)) \right) \right), \\ y_1 &= y_2, \quad \Omega = \omega_1, \quad 2\Omega = \omega_2 \end{aligned} \quad (57)$$

One may solve the above equations using the Runge-Kutta method to obtain a time response that is shown in Fig. 3. This response is periodic (repeats with a certain period  $T$ ) because the ratio between two frequencies is a rational number  $\omega_2 = 2\omega_1$ . If one plot the phase portrait by sampling at a time interval, it will yield a single point as shown in Fig. 3(c). It may be noted that the actual phase portrait is shown in Fig. 3(b). The Poincare' section of the periodic response is shown in Fig. 3(c). So, in this way, one can determine the Poincare' section by sampling the time response with the minimum period. It is the method used to reduce the dimension of the system by one. Let us take another case. Here, let excited voltage, i.e.,  $V(t) = (V_{DC} + V_{AC}^1 \cos(\omega_1 t) + V_{AC}^2 \cos(\omega_2 t))$  with two forcing terms having incommensurable frequencies. In this case, the ratio  $\omega_2 = 2\sqrt{2}\omega_1$  is an irrational number so, the response will be quasi-periodic. The time response is shown in Fig. 4(a). Here the system has more one period. By taking the sampling time as  $T$  (which is the minimum period), one may plot the phase portrait as shown in Fig. 4(b). Hence, for the quasi-periodic response, the Poincare' section is not a discrete point Fig. 4(c). It may be noted that in the case of incommensurable frequencies, one obtains multiple loops in the phase portraits and discrete points in the Poincare section.



**Fig. 3.** Time Response, Phase portrait, and Poincare' section,  $\beta = 0.1$ ,  $\Omega = \psi = \kappa = 1$ ,  $V_{AC} = 5$ ,  $V_{DC} = 20$ ,  $\xi = 4$ .



**Fig. 4.** Time Response, Phase portrait, and Poincare' section,  $\beta = 0.1$ ,  $V_{AC}^1 = V_{DC} = 100$ ,  $\psi = \kappa = \xi = 1$ ,  $V_{AC}^2 = 200$ ,  $\omega_2 = 12\sqrt{2}$ ,  $\omega_1 = 6$ .

### 279 5.3. Static case

280 By solving Eq. (56) moreover, neglecting time derivatives, i.e.,  $(\ddot{R}(t) = 0)$ , one can plot maximum static deflection of the  
281 micro-plate at  $w_{\max}(x, y) = w(0, 0)$  under electrostatic load.

282 Fig. 5a shows the variation of the maximum deflection of the micro-plate with respect to the electrostatic voltage for  
283 different values of  $\kappa$ . It can be seen that in the case  $\kappa = 1$ ,  $\xi = 0.2$  the pull-in voltage is equal to  $V_{DC}|_{Pull-in} = 2.5808$ . In this  
284 point  $\frac{\partial w_{\max}}{\partial V_{DC}}|_{V_{DC}(Pull-in)} = \infty$ . In other words, in the pull-in phenomenon, for a small change in voltage, there is a significant  
285 jump in the displacement response. After pull-in phenomena, the system enters the unstable state. With increasing electro-  
286 static force, the maximum deflection increases. This behavior is linear for small amounts of electrostatic force. By increasing  
287 the electrostatic force, this trend tends to highly nonlinear behavior. Also, by an increasing amount of  $\kappa$ , the pull-in voltage  
288 and maximum deflection will be increased. As seen in this figure, by increasing  $\kappa$ , the stiffness of the micro-plate increases,  
289 because, for a specific DC voltage, the maximum deflection of the micro-plate decrease.

290 Fig. 5b shows the variation of the maximum deflection of the micro-plate with respect to the electrostatic voltage for  
291 different values of  $\psi$ . As can be seen, with increasing  $\psi$ , the critical pull-in voltage, increases. As a general result, it can be  
292 said that with increasing  $\psi$ , the stiffness of the micro-plate increases.

#### 293 5.3.1. A semi-analytical solution for the transient response of micro-plate

294 The purpose of this section is to obtain a semi-analytical relationship for the static deflection of micro-plate under elec-  
295 trostatic load by the method of multiple scales. The semi-analytical relationship provides the possibility for analysis of the  
296 effects of different parameters analytically. Though the method of multiple scales is a solving method for obtaining the



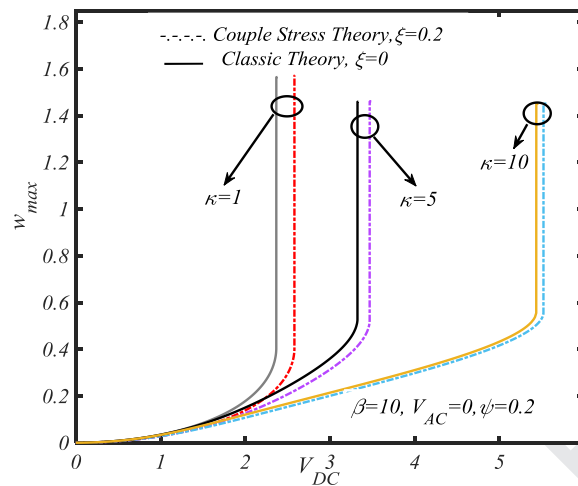


Fig. 5a. Static maximum deflection obtained for a micro-plate for different values of  $\kappa$ .

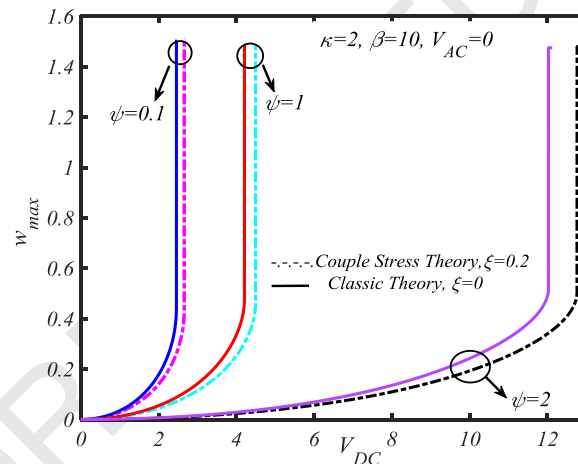


Fig. 5b. Variation of the maximum deflection of the micro-plate with respect to the electrostatic voltage for different values of  $\psi$ .

dynamic response of vibrating systems, here it is used to get the static response of the system. Approximate solution of Eq. (56) as a second-order expansion in terms of the positive and small parameter  $\varepsilon$  is as follows:

$$R(\tau_0, \tau_1, \tau_2, \varepsilon) = R_0(\tau_0, \tau_1, \tau_2) + \varepsilon R_1(\tau_0, \tau_1, \tau_2) + \varepsilon^2 R_2(\tau_0, \tau_1, \tau_2) \quad (58)$$

Where in the above equation  $R_0$ ,  $R_1$  and  $R_2$  are three unknown functions. To obtain a second-order uniform expansion by using the method of multiple scales, we need the three-time scales  $\tau_0$ ,  $\tau_1$ , and  $\tau_2$  in which are as follow:

$$\tau_0 = t, \tau_1 = \varepsilon t, \tau_2 = \varepsilon^2 t \quad (59)$$

In terms of the time scales  $\tau_i$ ,  $i = 0, 1, 2$ , the time derivatives become:

$$\frac{d}{dt} = D_0 + \varepsilon D_1 + \varepsilon^2 D_2 \quad \frac{d^2}{dt^2} = D_0^2 + 2\varepsilon D_0 D_1 + 2\varepsilon^2 D_0 D_2 + \varepsilon^2 D_1^2 \quad (60)$$

in which:

$$D_0 = \frac{d}{d\tau_0}, D_1 = \frac{d}{d\tau_1}, D_2 = \frac{d}{d\tau_2} \quad (61)$$

Using the timescales,  $\tau_i$ ,  $i = 0, 1, 2$  we transform Eq. (56) from an ordinary-differential equation into a partial differential equation:

$$\ddot{R}(t) + A_3 R(t) + \varepsilon^2 (A_1 R(t) \ddot{R}(t) + A_2 R(t)^2 \dot{R}(t) + A_4 R(t)^2 + A_5 R(t)^3 + A_6 R(t)^4 + A_7 R(t)^5) + A_8 = 0 \quad (62)$$

305 Equating coefficients of like powers of  $\varepsilon^0, \varepsilon^1, \varepsilon^2$  in Eq. (62):

$$O(\varepsilon^0) : D_0^2 R_0 + A_3 R_0 = -A_8 \quad (63)$$

306

$$O(\varepsilon^1) : D_0^2 R_1 + A_3 R_1 = -2D_0 D_1 R_0 \quad (63b)$$

307

$$O(\varepsilon^2) : D_0^2 R_2 + A_3 R_2 = -2D_0 D_1 R_1 - 2D_0 D_2 R_0 - D_1^2 R_0 - A_1 R_0 D_0^2 R_0 - A_2 R_0^2 D_0^2 R_0 - A_4 R_0^2 - A_5 R_0^3 - A_6 R_0^4 - A_7 R_0^5 \quad (64c)$$

308 The general solution for the Eq. (63) is as follows:

$$R_0 = -\frac{A_8}{A_3} + A(\tau_1, \tau_2) e^{i\sqrt{A_3}\tau_0} + \bar{A}(\tau_1, \tau_2) e^{-i\sqrt{A_3}\tau_0} \quad (64)$$

309 where  $A(\tau_1, \tau_2), \bar{A}(\tau_1, \tau_2)$  are the Complex Conjugate functions. By replacing Eq. (64) in (63) following equation is obtained:

310

$$D_0^2 R_1 + A_3 R_1 = -2i\sqrt{A_3} D_1 A(\tau_1, \tau_2) e^{i\sqrt{A_3}\tau_0} + 2i\sqrt{A_3} D_1 \bar{A}(\tau_1, \tau_2) e^{-i\sqrt{A_3}\tau_0} \quad (65)$$

311 Clearly, Eq. (65), breaks down because it contains secular terms and small-divisor terms. Due to  $R_1$  be periodic, we need  
312 to eliminate the secular and small-divisor terms. Therefore, we set the coefficient  $e^{\pm i\sqrt{A_3}\tau_0}$  equal to zero ( $D_1 A(\tau_1, \tau_2) =$   
313  $D_1 \bar{A}(\tau_1, \tau_2) = 0$ ). This result means that  $A(\tau_1, \tau_2)$  &  $\bar{A}(\tau_1, \tau_2)$  are only functions of  $\tau_2$ . Solving Eq. (65) gives Eq. (66) as fol-  
314 lows:

$$R_1 = B(\tau_1, \tau_2) e^{i\sqrt{A_3}\tau_0} + \bar{B}(\tau_1, \tau_2) e^{-i\sqrt{A_3}\tau_0} \quad (66)$$

315 Where  $B(\tau_1, \tau_2), \bar{B}(\tau_1, \tau_2)$  are Complex Conjugate. By replacing the Eqs. (64) and (66) into Eq. (63);

$$D_0^2 R_2 = CC + O.H.T + e^{i\omega_0 t} \times \left( -2i\sqrt{A_3} D_1 B - 2i\sqrt{A_3} D_2 A - 3A_5 A^2 \bar{A} - A_1 A_8 A V_p^2 + \frac{A_2 A_8^2 A}{A_3} + 3A_2 A_3 A^2 \bar{A} + \frac{2A_4 A_8 A}{A_3} - \frac{3A_5 A_8^2 A}{A_3^2} - \frac{5A_7 A_8^4 A}{A_3^4} - \frac{30A_7 A_8^2 A^2 \bar{A}}{A_3^2} + \frac{4A_6 A_8^3 A}{A_3^3} + \frac{12A_6 A_8 A^2 \bar{A}}{A_3} - 10A_7 A^3 \bar{A}^2 \right) \quad (67)$$

316 where  $CC$  &  $O.H.T$  are conjugate and other harmonic terms that are neglected in the further calculations. The elimination of  
317 the secular expressions in the above equation requires that the right side of the Eq. (67) be equal to zero except for the  
318 O.H.T terms. It follows that  $B$  is a function of  $\tau_2$ .

$$-2i\sqrt{A_3} D_2 A - 3A_5 A^2 \bar{A} - A_1 A_8 A + \frac{A_2 A_8^2 A}{A_3} + 3A_2 A_3 A^2 \bar{A} + \frac{2A_4 A_8 A}{A_3} - \frac{3A_5 A_8^2 A}{A_3^2} - \frac{5A_7 A_8^4 A}{A_3^4} - \frac{30A_7 A_8^2 A^2 \bar{A}}{A_3^2} + \frac{4A_6 A_8^3 A}{A_3^3} + \frac{12A_6 A_8 A^2 \bar{A}}{A_3} - 10A_7 A^3 \bar{A}^2 = 0 \quad (68)$$

319 It is appropriate that  $A$ , expressed as a polar state:

$$A = \frac{1}{2} a(\tau_2) e^{i\beta(\tau_2)} \quad (69)$$

320 where  $a$  and  $\beta$  are the real function of  $\tau_2$ . Substituting (90) into (89) and separating the result into real and imaginary parts,  
321 we obtain:

$$\dot{a}(\tau_2) = 0, \quad (70)$$

322

$$\dot{\beta} = \frac{\tau_2}{-\sqrt{A_3} a} \times \left( -\frac{3}{8} A_5 a^3 - \frac{1}{2} a A_1 A_8 + \frac{A_2 A_8^2 a}{2A_3} + \frac{3}{8} a^3 A_2 A_3 + \frac{A_4 A_8 a}{A_3} - \frac{3A_5 A_8^2 a}{2A_3^2} - \frac{5A_7 A_8^4 a}{2A_3^4} - \frac{30A_7 A_8^2 a^3}{8A_3^2} + \frac{2A_6 A_8^3 a}{A_3^3} + \frac{12A_6 A_8 a^3}{8A_3} - \frac{10}{32} A_7 a^5 \right) \quad (71)$$

323 where the dot ( $\dot{\cdot}$ ) denotes the derivative concerning to  $\tau_2$ . As  $\dot{a}(\tau_2) = 0$  therefore  $a$  is a constant and

$$a(\tau_2) = a_0, \quad (72)$$

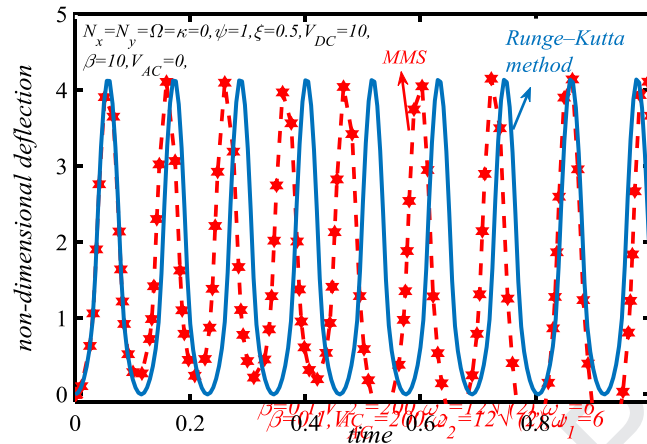


Fig. 6. Response of the micro-plate.

$$\beta = \frac{1}{-\sqrt{A_3}a} \times \left( -\frac{3}{8}A_5a^3 - \frac{1}{2}aA_1A_8 + \frac{A_2A_8^2a}{2A_3} + \frac{3}{8}a^3A_2A_3 + \frac{A_4A_8a}{A_3} - \frac{3A_5A_8^2a}{2A_3^2} - \frac{5A_7A_8^4a}{2A_3^4} - \frac{30A_7A_8^2a^3}{8A_3^2} + \frac{2A_6A_8^3a}{A_3^3} + \frac{12A_6A_8a^3}{8A_3} - \frac{10}{32}A_7a^5 \right) \tau_2 + \beta_0 \quad (73)$$

Here  $\beta_0$  is a constant. Now using  $\tau_2 = \varepsilon^2 t$  we reach to:

$$A = \frac{1}{2}a_0 \times \exp \left( i \frac{1}{-\sqrt{A_3}a} \left( -\frac{3}{8}A_5a^3 - \frac{1}{2}aA_1A_8 + \frac{A_2A_8^2a}{2A_3} + \frac{3}{8}a^3A_2A_3 + \frac{A_4A_8a}{A_3} - \frac{3A_5A_8^2a}{2A_3^2} - \frac{5A_7A_8^4a}{2A_3^4} - \frac{30A_7A_8^2a^3}{8A_3^2} + \frac{2A_6A_8^3a}{A_3^3} + \frac{12A_6A_8a^3}{8A_3} - \frac{10}{32}A_7a^5 \right) \varepsilon^2 t + \beta_0 i \right) \quad (74)$$

Substituting Eq. (74) in the expressions for  $R_0$ , i.e., Eq. (64) one obtains:

$$R_0 = -\frac{A_8}{A_3} + a_0 \times \cos \left( \sqrt{A_3} \tau_0 + \frac{1}{-\sqrt{A_3}a} \left( -\frac{3}{8}A_5a^3 - \frac{1}{2}aA_1A_8 + \frac{A_2A_8^2a}{2A_3} + \frac{3}{8}a^3A_2A_3 + \frac{A_4A_8a}{A_3} + \frac{3}{8}a^3A_2A_3 + \frac{A_4A_8a}{A_3} - \frac{3A_5A_8^2a}{2A_3^2} - \frac{5A_7A_8^4a}{2A_3^4} - \frac{30A_7A_8^2a^3}{8A_3^2} + \frac{2A_6A_8^3a}{A_3^3} + \frac{12A_6A_8a^3}{8A_3} - \frac{10}{32}A_7a^5 \right) \tau_2 + \beta_0 \right) \quad (75)$$

By replacing  $t = 0$  in the relation (75), the equation of the static deflection and with derivative from the relation (75), and setting  $t = 0$ , the relation of the initial velocity is obtained. Now, if the time tends to the infinitely, in other words  $\tau_i \rightarrow \infty$ ,  $i = 1, 2$ , Eq. (75) only has a mathematical meaning that  $a_0$  goes to zero ( $a_0 \rightarrow 0$ ). The remainder relationship is independent of time, indicating the maximum deflection of the micro-plate.

$$w_{\max} = -\frac{A_8}{A_3} \quad (76)$$

Fig. 6 shows the comparison between the response of Multiple scales method (MMS) and Runge-Kutta method with displacement and initial velocity of zero. As can be seen, the results of non-dimensional deflection obtained by Runge-kutta method with respect to time are in good agreement with the findings results presented by MMS.

#### 5.4. Dynamic response of rectangular micro-plate under electrical excitation

In this section, the dynamic response of rectangular micro-plate under electrical excitation, consisting of DC constant and alternating AC voltages are studied. DC Voltage component causes to bend micro-plate to a new position. Then the AC voltage causes vibration the micro-plate around the new equilibrium point. Such structures are used in resonator microscopes. Obtaining a frequency response curve in primary and secondary resonance modes and investigating instability dynamics

(pull-in) are the goals of this section. First, the system's response to the electric voltage in the primary resonance is studied by the method of multiple scales. In the following, the dynamic behavior of the micro-plate under Subharmonic and Superharmonic resonance conditions are also studied. In this section, it is supposed that the micro-plate shown in Fig. 1 is excited by an electrical voltage  $V(t) = V_{DC} + V_{AC} \cos(\Omega t)$ , where  $V_{DC}$  is the constant voltage,  $V_{AC}$  is the amplitude of variable voltage and  $\Omega$  is the excitation frequency.

#### 5.4.1. Primary resonance using multiple scale method considering a weak forcing function

In the primary resonance state, the excitation force frequency is very close to the fundamental natural frequency of the system. As discussed in the previous section, in the absence of external force, the amplitude of free vibration response is a function of the natural frequency ( $\omega_0$ ). Similar to the linear vibration here we may consider the behavior of the micro-plate near the resonance condition, i.e., when the external frequency is equal to the natural frequency of the system. This condition is known as the primary resonance condition ( $\Omega \approx \omega_0$ ). To study the behavior of the system near the primary resonance condition, one may use the detuning parameter which represents the nearness of the external frequency to that of the natural frequency. Hence one may write:

$$\Omega = \omega_0 + \varepsilon^2 \sigma \quad (77)$$

where  $\Omega$  is the excitation frequency and  $\omega_0$  is the linear natural frequency of the system.  $\sigma$  is the detuning parameter that indicates the proximity of the excited frequency to the linear frequency of the system. In the primary resonance condition, the amplitude of the excitation force is considered the same order as the nonlinear terms. Therefore, the excitation voltage is sorted as follows.

$$V(t) = V_{DC} + \varepsilon^2 V_{AC} \cos(\Omega t) \quad (78)$$

As a result, Eq. (56) found the form as follows:

$$\begin{aligned} \ddot{R}(t) + A_3 R(t) + A_9 (V_{DC} + \varepsilon^2 V_{AC} \cos(\Omega t))^2 + \varepsilon^2 (A_1 R(t) \dot{R}(t) \\ + A_2 R(t)^2 \ddot{R}(t) + A_4 R(t)^3 + A_5 R(t)^3 + A_6 R(t)^4 + A_7 R(t)^5) = 0, \\ A_9 = -\frac{16\beta}{9} \end{aligned} \quad (79)$$

In this analysis, it is assumed that  $V_{DC}^2 \gg V_{AC}^2$ . Substituting Eq. (60) in Eq. (79) moreover, separate terms with a different order of  $\varepsilon$ , one obtains the following equations.

$$O(\varepsilon^0) : D_0^2 R_0 + A_3 R_0 = -A_9 V_{DC}^2 \quad (80)$$

$$O(\varepsilon^1) : D_0^2 R_1 + A_3 R_1 = -2D_0 D_1 R_0 \quad (81)$$

$$\begin{aligned} O(\varepsilon^2) : D_0^2 R_2 + A_3 R_2 = -2D_0 D_1 R_1 - 2D_0 D_2 R_0 - D_1^2 R_0 \\ - A_1 R_0 D_0^2 R_0 - A_2 R_0^2 D_0^2 R_0 - A_4 R_0^2 - A_5 R_0^3 - A_6 R_0^4 - A_7 R_0^5 \\ - 2A_9 V_{AC} V_{DC} \cos(\Omega t) \end{aligned} \quad (82)$$

The general solution for the Eq. (80) is as follows:

$$R_0 = -\frac{A_9 V_{DC}^2}{A_3} + A(\tau_1, \tau_2) e^{i\sqrt{A_3}\tau_0} + \bar{A}(\tau_1, \tau_2) e^{-i\sqrt{A_3}\tau_0} \quad (83)$$

By substituting Eq. (83) in Eq. (81) one obtains the following equation.

$$D_0^2 R_1 + A_3 R_1 = -2i\sqrt{A_3} D_1 A(\tau_1, \tau_2) e^{i\sqrt{A_3}\tau_0} + 2i\sqrt{A_3} D_1 \bar{A}(\tau_1, \tau_2) e^{-i\sqrt{A_3}\tau_0} \quad (84)$$

By solving Eq. (84), one can write

$$R_1 = B(\tau_1, \tau_2) e^{i\sqrt{A_3}\tau_0} + \bar{B}(\tau_1, \tau_2) e^{-i\sqrt{A_3}\tau_0} \quad (85)$$

By replacing (83), (85) and (77) into the Eq. (82); the following equation is obtained.

$$\begin{aligned} D_0^2 R_2 + A_3 R_2 = CC + O.H.T + e^{i\omega_0 t} \times \left( -2i\sqrt{A_3} D_1 B - 2i\sqrt{A_3} D_2 A - 3A_5 A^2 \bar{A} \right. \\ \left. - A_1 A_9 A V_{DC}^2 + \frac{A_2 A_9^2 V_{DC}^2 A}{A_3} + 3A_2 A_3 A^2 \bar{A} + \frac{2A_4 A_9 V_{DC}^2 A}{A_3} - \frac{3A_5 A_9^2 V_{DC}^4 A}{A_3^2} \right. \\ \left. - \frac{5A_7 A_9^4 V_{DC}^8 A}{A_3^4} - \frac{30A_7 A_9^2 V_{DC}^4 A^2 \bar{A}}{A_3^2} + \frac{4A_6 A_9^3 A V_{DC}^6}{A_3^3} + \frac{12A_6 A_9 V_{DC}^2 A^2 \bar{A}}{A_3} \right) \end{aligned}$$

$$-10A_7A^3\bar{A}^2 - A_9V_{AC}V_{DC}e^{i\sigma\tau_2}) \quad (86)$$

To eliminate the secular and near secular terms from Eq. (86), one can write:

$$\begin{aligned} & -2i\sqrt{A_3}D_2A - 3A_5A^2\bar{A} - A_1A_9AV_{DC}^2 + \frac{A_2A_9^2V_{DC}^2A}{A_3} + 3A_2A_3A^2\bar{A} + \frac{2A_4A_9V_{DC}^2A}{A_3} \\ & - \frac{3A_5A_9^4V_{DC}^8A}{A_3^4} - \frac{5A_7A_9^4V_{DC}^8A}{A_3^4} - \frac{30A_7A_9^2V_{DC}^4A^2\bar{A}}{A_3^2} + \frac{4A_6A_9^3V_{DC}^6A}{A_3^3} \\ & + \frac{12A_6A_9V_{DC}^2A^2\bar{A}}{A_3} - 10A_7A^3\bar{A}^2 - A_9V_{AC}V_{DC}e^{i\sigma\tau_2} = 0 \end{aligned} \quad (87)$$

It follows that  $B$  is only a function of  $\tau_2$ . Now by substituting  $A = \frac{1}{2}a(\tau_2)e^{i\beta(\tau_2)}$  in Eq. (87) moreover, separating the real and imaginary parts, following reduced equations are obtained.

$$\dot{a} = \frac{-A_9V_{AC}V_{DC}}{\sqrt{A_3}} \sin(\sigma\tau_2 - \beta), \quad (88)$$

$$\begin{aligned} a\dot{\beta} = & \frac{1}{-\sqrt{A_3}} \left\{ \left( -\frac{1}{2}A_1A_9V_{DC}^2 + \frac{1}{2}\frac{A_2A_9^2V_{DC}^2}{A_3} + \frac{A_4A_9V_{DC}^2}{A_3} - \frac{3}{2}\frac{A_5A_9^2V_{DC}^4}{A_3^2} - \right. \right. \\ & \left. \frac{5}{2}\frac{A_7A_9^4V_{DC}^8}{A_3^4} + \frac{2A_6A_9^3V_{DC}^6}{A_3^3} \right) a + \left( -\frac{3}{8}A_5 - \frac{30}{8}\frac{A_7A_9^2V_{DC}^4}{A_3^2} + \right. \\ & \left. \frac{12}{8}\frac{A_6A_9V_{DC}^2}{A_3} + \frac{3}{8}A_2A_3 \right) a^3 - \frac{10}{32}A_7a^5 - A_9V_{AC}V_{DC} \cos(\sigma\tau_2 - \beta) \left. \right\} \end{aligned} \quad (89)$$

To write these two equations in its autonomous form one may use  $\gamma = \sigma\tau_2 - \beta$  and obtained the following equations.

$$\dot{a} = \frac{-A_9V_{AC}V_{DC}}{\sqrt{A_3}} \sin(\gamma), \quad (90)$$

$$\begin{aligned} a\dot{\gamma} = & a\sigma + \frac{1}{\sqrt{A_3}} \left( \left( -\frac{1}{2}A_1A_9V_{DC}^2 + \frac{1}{2}\frac{A_2A_9^2V_{DC}^2}{A_3} + \frac{A_4A_9V_{DC}^2}{A_3} - \frac{3}{2}\frac{A_5A_9^2V_{DC}^4}{A_3^2} - \right. \right. \\ & \left. \frac{5}{2}\frac{A_7A_9^4V_{DC}^8}{A_3^4} + \frac{2A_6A_9^3V_{DC}^6}{A_3^3} \right) a + \left( -\frac{3}{8}A_5 - \frac{30}{8}\frac{A_7A_9^2V_{DC}^4}{A_3^2} + \right. \\ & \left. \frac{12}{8}\frac{A_6A_9V_{DC}^2}{A_3} + \frac{3}{8}A_2A_3 \right) a^3 - \frac{10}{32}A_7a^5 - A_9V_{AC}V_{DC} \cos(\gamma) \left. \right) \end{aligned} \quad (91)$$

One should solve these two equations to obtain  $a$  and  $\gamma$ . Now for the steady state as  $\dot{a}$  and  $\dot{\gamma}$  equals to 0, one can write Eqs. (90) and (91) as:

$$\frac{-A_9V_{AC}V_{DC}}{\sqrt{A_3}} \sin(\gamma) = 0, \quad (92)$$

$$\begin{aligned} a\sqrt{A_3}\sigma + & \left( -\frac{1}{2}A_1A_9V_{DC}^2 + \frac{1}{2}\frac{A_2A_9^2V_{DC}^2}{A_3} + \frac{A_4A_9V_{DC}^2}{A_3} - \frac{3}{2}\frac{A_5A_9^2V_{DC}^4}{A_3^2} - \right. \\ & \left. \frac{5}{2}\frac{A_7A_9^4V_{DC}^8}{A_3^4} + \frac{2A_6A_9^3V_{DC}^6}{A_3^3} \right) a + \left( -\frac{3}{8}A_5 - \frac{30}{8}\frac{A_7A_9^2V_{DC}^4}{A_3^2} + \right. \\ & \left. \frac{12}{8}\frac{A_6A_9V_{DC}^2}{A_3} + \frac{3}{8}A_2A_3 \right) a^3 - \frac{10}{32}A_7a^5 = A_9V_{AC}V_{DC} \cos(\gamma) \end{aligned} \quad (93)$$

Now eliminating  $\gamma$  from the above equations, one obtains:

$$\begin{aligned} \sigma = & \frac{1}{a\sqrt{A_3}} \times \left( \pm A_9V_{AC}V_{DC} - \left( \left( -\frac{1}{2}A_1A_9V_{DC}^2 + \frac{1}{2}\frac{A_2A_9^2V_{DC}^2}{A_3} + \frac{A_4A_9V_{DC}^2}{A_3} - \right. \right. \right. \\ & \left. \left. \frac{3}{2}\frac{A_5A_9^2V_{DC}^4}{A_3^2} - \frac{5}{2}\frac{A_7A_9^4V_{DC}^8}{A_3^4} \right) a + \left( -\frac{3}{8}A_5 - \frac{30}{8}\frac{A_7A_9^2V_{DC}^4}{A_3^2} + \frac{12}{8}\frac{A_6A_9V_{DC}^2}{A_3} \right. \right. \end{aligned}$$

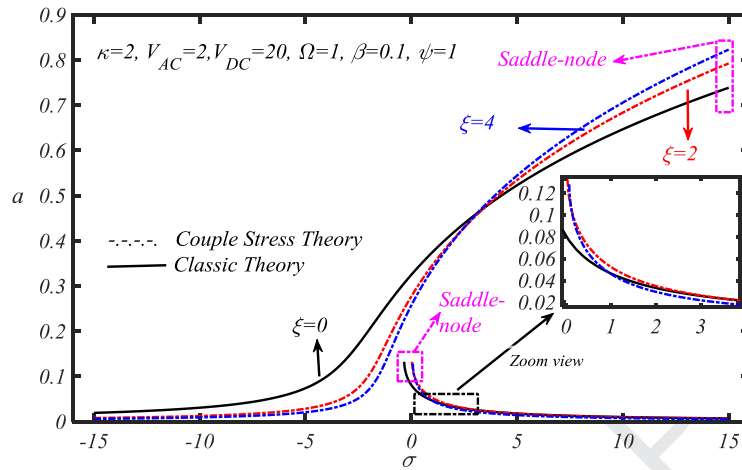


Fig. 7. Frequency response curve of a rectangular micro-plate for different values of  $\xi$ .

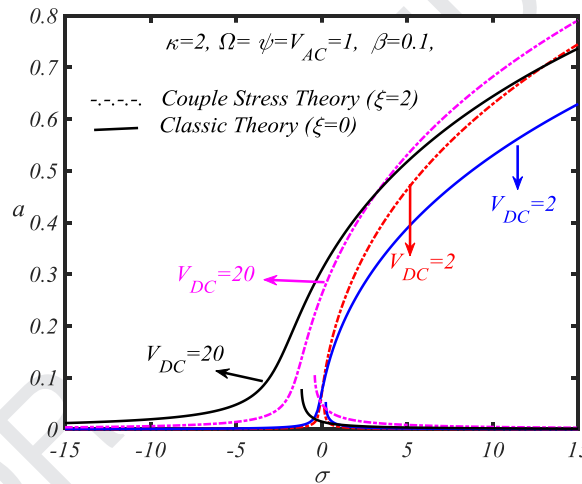


Fig. 8. Frequency response curve of a rectangular micro-plate for different values of  $V_{DC}$ .

$$\left. + \frac{3}{8} A_2 A_3 \right) a^3 + \frac{4 A_6 A_9^3 V_{DC}^6}{A_3^3} - \frac{10}{32} A_7 a^5 \Bigg) \quad (94)$$

375 The frequency response curve of a rectangular micro-plate for  $V_{DC} = 20$ ,  $V_{AC} = 2$  is given in Fig. 7.

376 As can be seen, the existence of nonlinear terms in the governing equation leads to appearance plate stiffness property.  
 377 The saddle-node bifurcation points have been shown in Fig. 7. To identify the pull-in dynamic voltage, the AC voltage re-  
 378 sponse should be calculated. For this purpose, the amplitude range of the vibration should be plotted in terms of the excited  
 379 amplitude. With increasing the voltage when the excitation voltage reaches to pull-in dynamic voltage, a sudden change in  
 380 the amplitude of the vibration is observed. This sudden change in the vibration amplitude can be so high, that leads to a  
 381 collision of a capacitor micro-plate with the fixed electrode. Therefore, for the design of such systems, it is necessary to  
 382 care about critical voltage. It should be noted that, due to assumption ( $V_{DC}^2 \gg V_{AC}^2$ ), so by increasing AC voltage, this as-  
 383 sumption can cause many errors in the semi-analytical response. However, in a low value of  $V_{AC}$  the proposed method is  
 384 an appropriate method. Fig. 8 shows the frequency response curve of a rectangular micro-plate for different DC voltages.  
 385 According to this figure with increasing constant voltage ( $V_{DC}$ ) stiffness of the micro-plate becomes less.

386 Fig. 9 shows the frequency response curve of a rectangular micro-plate for a DC voltage  $V_{DC} = 20$  (Volt) and different AC  
 387 voltages.

388 Regarding Fig. 9, it can be seen that with increasing the alternating voltage,  $V_{AC}$  the branches of each curve get more  
 389 spaced apart. In other words, the voltages,  $V_{AC}$ , does not have any effect on the level of stiffness of the micro-plate and only  
 390 acts as an external excitation dynamic force. Fig. 10 shows the frequency response curve of a rectangular micro-plate for the  
 391 AC voltage  $V_{AC} = 2$  volt and DC voltage  $V_{DC} = 20$  volt, for the different value of  $\kappa$ . According to this figure, it can be seen that  
 392 the reduction of the non-dimensional parameter  $\kappa$  reduces the stiffness of the micro-plate as it has already been mentioned



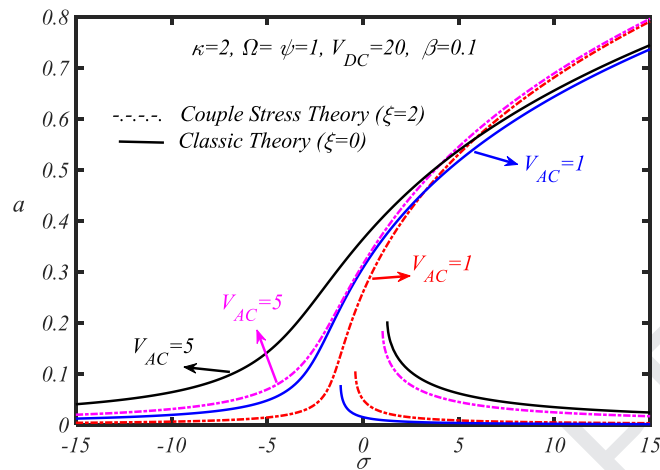


Fig. 9. Frequency response curve of a rectangular micro-plate for different values of  $V_{AC}$ .

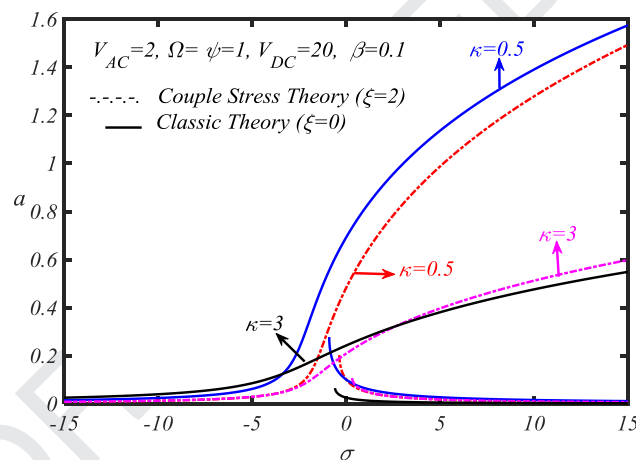


Fig. 10. Frequency response curve of a rectangular micro-plate for different values of  $\kappa$ .

for the static case (Fig. 5). In other words, reducing  $\kappa$  reduces the effect of the stretch of the mid-plane and the stiffness of the micro-plate.

Fig. 11 indicates the frequency response curve of a rectangular micro-plate for different values of  $\psi$ . According to this figure, a significant change in the frequency response curve is observed with decreasing  $\psi$  from 3 to 1.

Fig. 12 shows the vibration amplitude curve according to the amplitude of the excitation force for a rectangular micro-plate for different values of  $\sigma$ . As seen it turns out that for positive values of  $\sigma$  for a given alternating voltage, there are several values for the vibration amplitude. However, for negative values of  $\sigma$  for an alternating voltage, there are not multiple values for the vibration amplitude. On the other hand, for positive values of  $\sigma$  by increasing  $\sigma$ , an increase in the critical dynamic voltage occurs. For instance, in the two cases of modified couple stress theory, pull-in voltages are as follow:

$\sigma = 3$ ;  $\{a = 0.222$  and,  $\sigma = 1$ ;  $\{a = 0.138$  As a significant result, it can be stated that for a specific DC voltage, the critical dynamic voltage is a function of the forced excitation frequency. The effect of the mid-plane stretching on critical dynamic voltage is now investigated in Fig. 13.

This figure shows vibration amplitude in terms of the excitation force amplitude for different values of  $\kappa$ . As can be seen, with increasing  $\kappa$ , the critical dynamic voltage decreases. Previously mentioned that  $\kappa$  is the value of the stretch of the mid-plane. Therefore, for a given excitation frequency ( $\Omega = 1$ ), the increase in the  $\kappa$  decreases the critical dynamic voltage.

Fig. 14 shows the vibration amplitude curve according to the amplitude of the excitation force for different values of  $\psi$ . As can be seen, by increasing  $\psi$ , the critical dynamic voltage ( $V_{AC(Pull-in)}$ ) decreases.

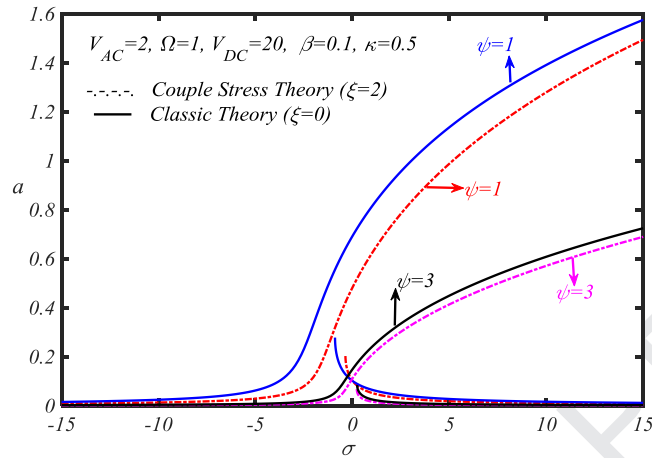


Fig. 11. Frequency response curve of a rectangular micro-plate for different values of  $\psi$ .

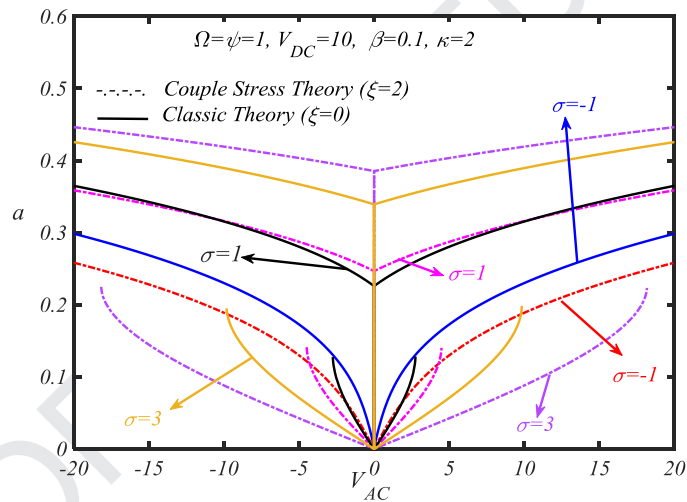


Fig. 12. Vibration amplitude curve of a rectangular micro-plate for different values of  $\sigma$ .

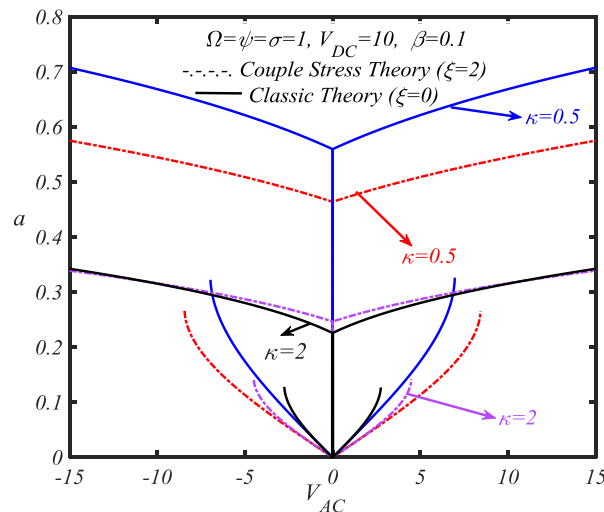


Fig. 13. Vibration amplitude in terms of the excitation force amplitude for a rectangular micro-plate for different values of  $\kappa$ .

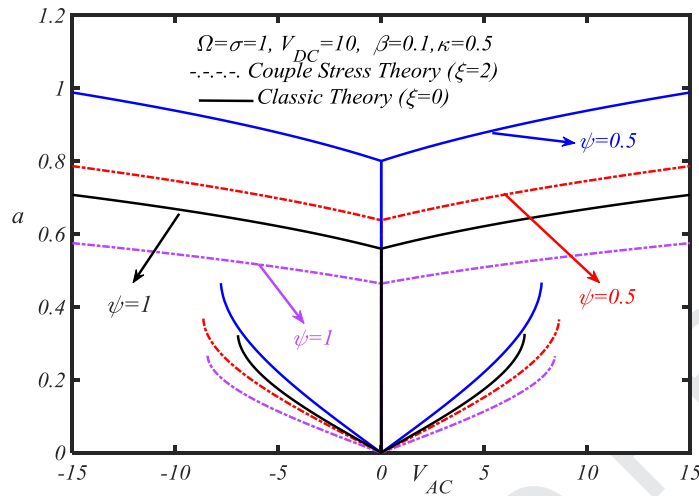


Fig. 14. Vibration amplitude in terms of the excitation force amplitude for a rectangular micro-plate for different values of  $\psi$ .

#### 5.4.2. Secondary resonance

In the secondary resonance, the domain of excitation load is not assumed to be small [62], i.e., the forcing term is assumed to be of same order as that of the linear term. So, the equation of motion considered in this case as:

$$\begin{aligned} \ddot{R}(t) + A_3 R(t) + \varepsilon (A_1 R(t) \ddot{R}(t) + A_2 R(t)^2 \ddot{R}(t) + A_4 R(t)^2 + A_5 R(t)^3 + A_6 R(t)^4 + A_7 R(t)^5) \\ = -A_9 (V_{DC} + V_{AC} \cos(\Omega t))^2 \end{aligned} \quad (95)$$

Using the method of multiple scales, the solution of Eq. (95) can be written as:

$$R(\tau_0, \tau_1, \varepsilon) = R_0(\tau_0, \tau_1) + \varepsilon R_1(\tau_0, \tau_1) \quad (96)$$

Now by separating the terms with a different order of  $\varepsilon$ , one obtains the following equations:

$$O(\varepsilon^0) : D_0^2 R_0 + A_3 R_0 = -A_9 (V_{DC} + V_{AC} \cos(\Omega t))^2 \quad (97)$$

$$\begin{aligned} O(\varepsilon^1) : D_0^2 R_1 + A_3 R_1 = -2D_0 D_1 R_0 - A_1 R_0 D_0^2 R_0 - A_2 Z_0^2 D_0^2 R_0 \\ - A_4 R_0^2 - A_5 R_0^3 - A_6 R_0^4 - A_7 R_0^5 \end{aligned} \quad (98)$$

The solution of Eq. (97) can be written as:

$$R_0 = -\frac{A_9 V_{DC}^2}{2A_3} + A(\tau_1) e^{i\omega_0 \tau_0} + \Lambda e^{i\Omega \tau_0} + CC \quad (99)$$

where

$$\omega_0 = \sqrt{A_3}, \quad \Lambda = -\frac{A_9 V_{DC} V_{AC}}{(\omega_0^2 - \Omega^2)} \quad (100)$$

It may be noted that unlike the previous section, where only the complementary part of the solution was present, in this case, both complimentary and particular integral parts, are present in the solution of  $R_0$ . It may be noted that when the exponent terms of the mixed secular terms are equal to  $\omega_0$  a resonance condition will occur. Hence, resonance will be

$$\begin{aligned} \Omega = \omega_0, & \quad (\text{Primary resonance}) \\ \Omega = 3\omega_0, & \quad (\text{Sub harmonic resonance}) \\ 3\Omega = \omega_0, & \quad (\text{Superharmonic Resonance}) \end{aligned}$$

#### 5.4.3. Superharmonic resonance

To express the nearness of the external excitation frequency to one-third of the natural frequency one may use the detuning parameter  $\Omega \approx \frac{1}{3}\omega_0$  as follows:

$$3\Omega = \omega_0 + \sigma \varepsilon \quad (101)$$

Now to eliminate the secular and near secular terms from Eq. (98) one can write:

$$\begin{aligned} -A_5 \Lambda^3 e^{i\varepsilon \sigma \tau_0} - 2i\omega_0 D_1 A + A_2 \Lambda^3 \Omega^2 e^{i\varepsilon \sigma \tau_0} - 6A_5 A \Lambda^2 - 3A_5 A^2 \bar{A} + 3A_2 A^2 \bar{A} \omega^2 \\ - \frac{10A_7 A_9^2 V_{DC}^4 \Lambda^3 e^{i\varepsilon \sigma \tau_0}}{A_3^2} - \frac{60A_7 A_9^2 V_{DC}^4 A \Lambda^2}{A_3^2} - \frac{3A_5 A_9^2 V_{DC}^4 A}{A_3^2} - 60A_7 A^2 \bar{A} \Lambda^2 \end{aligned}$$

$$\begin{aligned}
& + \frac{2A_4A_9V_{DC}^4A}{A_3} + \frac{12A_6A_9V_{DC}^2A^2\bar{A}}{A_3} + \frac{4A_6A_9V_{DC}^2\Lambda^3e^{i\varepsilon\sigma\tau_0}}{A_3} - 10A_7A^3\bar{A}^2 \\
& - \frac{5A_7A_9^4V_{DC}^8A}{A_3^4} - 30A_7A\Lambda^4 + \frac{24A_6A_9V_{DC}^2A\Lambda^2}{A_3} + 4A_2A\Lambda^2\Omega^2 \\
& + 2A_2A\Lambda^2\omega_0^2 + \frac{A_2A_9^2V_{DC}^4A\omega_0^2}{A_3^2} - 5A_7\Lambda^5e^{i\varepsilon\sigma\tau_0} = 0
\end{aligned} \quad (102)$$

Using  $A = \frac{1}{2}a(\tau_1)e^{i\beta(\tau_1)}$  and separating the real and imaginary parts following reduced equations are obtained.

$$\begin{aligned}
\dot{a} &= \frac{1}{\omega_0} \left( -A_5\Lambda^3 + A_2\Lambda^3\Omega^2 - \frac{10A_7A_9^2V_{DC}^4\Lambda^3}{A_3^2} + \frac{4A_6A_9V_{DC}^2\Lambda^3}{A_3} - 5A_7\Lambda^5 \right) \times \\
\sin(\sigma\tau_1 - \beta) &= 0,
\end{aligned} \quad (103)$$

$$\begin{aligned}
a\dot{\beta} &= \frac{\cos(\sigma\tau_1 - \beta)}{-\omega_0} \left( \left( -A_5\Lambda^3 + A_2\Lambda^3\Omega^2 - \frac{10A_7A_9^2V_{DC}^4\Lambda^3}{A_3^2} + \frac{4A_6A_9V_{DC}^2\Lambda^3}{A_3} - 5A_7\Lambda^5 \right) \right. \\
& - 3A_5a\Lambda^2 - \frac{3}{8}A_5a^3 \frac{30A_7A_9^2V_{DC}^4a\Lambda^2}{A_3^2} - \frac{3A_5A_9^2V_{DC}^4a}{2A_3^2} + \frac{3}{8}A_2a^3\omega_0^2 + \frac{A_4A_9V_{DC}^2a}{A_3} \\
& + \frac{12A_6A_9V_{DC}^2a^3}{8A_3} - \frac{60}{8}A_7a^3\Lambda^2 - \frac{5A_7A_9^4V_{DC}^8a}{2A_3^4} - \frac{10}{32}A_7a^5 - 15A_7a\Lambda^4 \\
& \left. + \frac{12A_6A_9V_{DC}^2a\Lambda^2}{A_3} + 2A_2a\Lambda^2\Omega^2 + A_2a\Lambda^2\omega_0^2 + \frac{A_2A_9^2V_{DC}^4a\omega_0^2}{2A_3^2} \right)
\end{aligned} \quad (104)$$

Now to express the above equations in their autonomous form one may use the following transformation.,  $\gamma = \sigma\tau_1 - \beta$

Hence, Eq. (103) can be written as:

$$\dot{a} = \frac{\sin(\gamma)}{\omega_0} \times \left( -A_5\Lambda^3 + A_2\Lambda^3\Omega^2 - \frac{10A_7A_9^2V_{DC}^4\Lambda^3}{A_3^2} + \frac{4A_6A_9V_{DC}^2\Lambda^3}{A_3} - 5A_7\Lambda^5 \right), \quad (105)$$

$$\begin{aligned}
a\dot{\gamma} &= a\sigma + \frac{1}{\omega_0} \left( \left( -A_5\Lambda^3 + A_2\Lambda^3\Omega^2 - \frac{10A_7A_9^2V_{DC}^4\Lambda^3}{A_3^2} + \frac{4A_6A_9V_{DC}^2\Lambda^3}{A_3} \right. \right. \\
& - 5A_7\Lambda^5 \left. \right) \cos(\sigma\tau_1 - \beta) - 3A_5a\Lambda^2 - \frac{3}{8}A_5a^3 \frac{30A_7A_9^2V_{DC}^4a\Lambda^2}{A_3^2} - \frac{3A_5A_9^2V_{DC}^4a}{2A_3^2} \\
& + \frac{3}{8}A_2a^3\omega_0^2 + \frac{A_4A_9V_{DC}^2a}{A_3} + \frac{12A_6A_9V_{DC}^2a^3}{8A_3} - \frac{60}{8}A_7a^3\Lambda^2 - \frac{5A_7A_9^4V_{DC}^8a}{2A_3^4} - \frac{10}{32}A_7a^5 \\
& \left. - 15A_7a\Lambda^4 + \frac{12A_6A_9V_{DC}^2a\Lambda^2}{A_3} + 2A_2a\Lambda^2\Omega^2 + A_2a\Lambda^2\omega_0^2 + \frac{A_2A_9^2V_{DC}^4a\omega_0^2}{2A_3^2} \right)
\end{aligned} \quad (106)$$

For steady state, the time derivative terms should be vanished, in Eq. (105) and (106), i.e.,  $\dot{a} = \dot{\gamma} = 0$ . Now by eliminating  $\gamma$  from the above two equations, one can obtain a closed-form equation which can be used for finding the frequency response of the system and the relation between the detuning parameter and the amplitude of the response as follows:

$$\begin{aligned}
\sigma &= -\frac{1}{a\omega_0} \left( \left( \pm \left( -A_5\Lambda^3 + A_2\Lambda^3\Omega^2 - \frac{10A_7A_9^2V_{DC}^4\Lambda^3}{A_3^2} + \frac{4A_6A_9V_{DC}^2\Lambda^3}{A_3} - 5A_7\Lambda^5 \right) \right. \right. \\
& + (-3A_5a\Lambda^2 - A_5a^3 \frac{90A_7A_9^2V_{DC}^4a\Lambda^2}{8A_3^2} - \frac{3A_5A_9^2V_{DC}^4a}{2A_3^2} + \frac{3}{8}A_2a^3\omega_0^2 + \frac{A_4A_9V_{DC}^2a}{A_3} \\
& + \frac{12A_6A_9V_{DC}^2a^3}{8A_3} - \frac{60A_7a^3\Lambda^2}{8} - \frac{5A_7A_9^4V_{DC}^8a}{2A_3^4} - \frac{10A_7a^5}{32} - 15A_7a\Lambda^4 \\
& \left. \left. + \frac{12A_6A_9V_{DC}^2a\Lambda^2}{A_3} + 2A_2a\Lambda^2\Omega^2 + A_2a\Lambda^2\omega_0^2 + \frac{A_2A_9^2V_{DC}^4a\omega_0^2}{2A_3^2} \right) \right)
\end{aligned} \quad (107)$$

Hence, in this resonance condition, the nonlinearity adjusts the frequency of the free oscillation term to precisely three times the frequency of the excitations so that the response is periodic. Since the frequency of the free oscillation term is three times the frequency of excitation, such resonances are called superharmonic resonances [62]. Fig. 15 shows the frequency response curve of a micro-plate under electric voltage under superharmonic resonance state for different values of  $\kappa$ .

According to the above figure, it can be seen that by increasing the values of  $\kappa$  result in decreases the region in the frequency response curve. Also, with changing  $\kappa$  in large values of  $\sigma$ , the slip of frequency response diagrams does not change. It should be noted that by increasing  $\kappa$  the backbone curve tends to be moved to the right side.

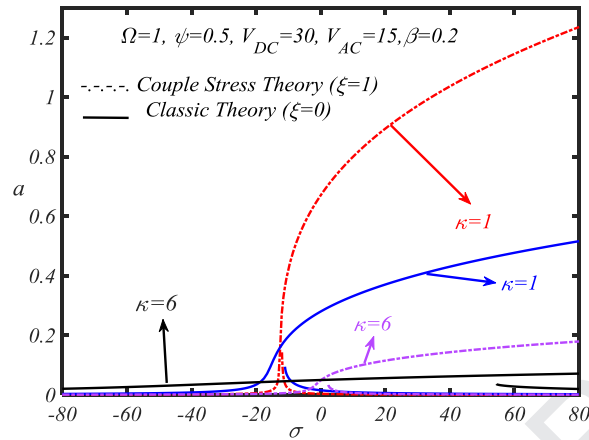


Fig. 15. Frequency response curve of a micro-plate under electric voltage and superharmonic resonance state for different values of  $\kappa$ .

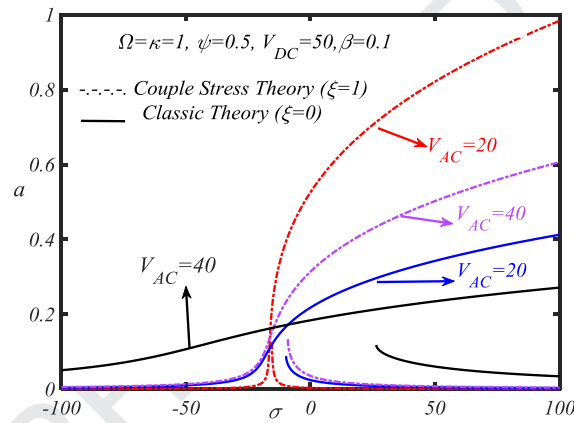


Fig. 16. Frequency response curve of a micro-plate under electric voltage and superharmonic resonance state for different values of  $V_{AC}$ .

Fig. 16. shows the frequency response curve of a micro-plate under electric voltage in superharmonic resonance state for different  $V_{AC}$  values. It can be seen that by increasing the values of  $V_{AC}$  result in decreases the region in the frequency response curve. Also, for any values of  $V_{AC}$ , the slip of frequency response diagrams does not change. This trend means that changing the AC voltage ( $V_{AC}$ ) does not affect the stiffness of the micro-plate.

Fig. 17. shows the frequency response curve of a micro-plate under electric voltage in superharmonic resonance state for different  $V_{DC}$  values. It should be noted that by increasing  $V_{DC}$  stiffness of the micro-plate reduced, and the backbone curve tends to be moved to the right side.

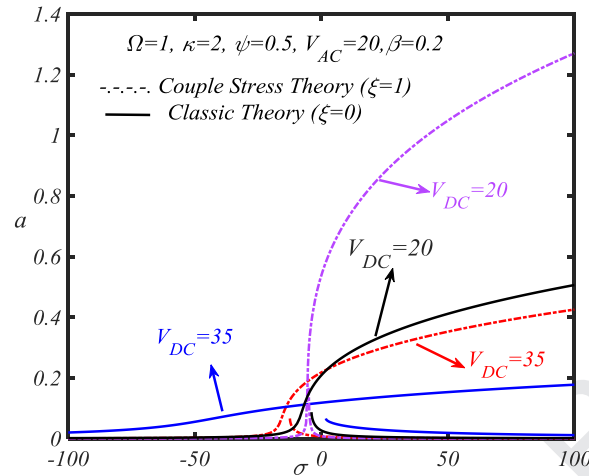
#### 5.4.4. Subharmonic resonance

When the external frequency is nearly three times the natural frequency of the system, using detuning parameter one can write

$$\Omega = 3\omega_0 + \sigma\epsilon \quad (108)$$

Using a similar procedure of the multiple scale method, eliminating the secular terms from Eq. (98) one can write:

$$\begin{aligned} & -3A_5A^2\bar{A} - 6A_5A\Lambda^2 - \frac{A_1A_9V_{DC}^2A\omega_0^2}{A_3} + 4A_2A\Lambda^2\Omega^2 + \frac{4A_6A_9^3V_{DC}^6A}{A_3^3} \\ & + \frac{12A_6A_9V_{DC}^2A^2\bar{A}}{A_3} + \frac{24A_6A_9V_{DC}^2A\Lambda^2}{A_3} + \frac{12A_6A_9V_{DC}^2\bar{A}^2\Lambda e^{i\epsilon\sigma\tau_0}}{A_3} + 2A_2\bar{A}^2\Lambda\omega_0^2e^{i\epsilon\sigma\tau_0} \\ & - 60A_7A^2\bar{A}\Lambda^2 - \frac{3A_5A_9^2V_{DC}^4A}{A_3^2} + \frac{2A_4A_9V_{DC}^2A}{A_3} + \frac{A_2A_9^2V_{DC}^4A\omega_0^2}{A_3^2} - 30A_7A\Lambda^4 \\ & - 10A_7A^3\bar{A}^2 - 2i\omega_0D_1A - 20A_7A\bar{A}^3\Lambda e^{i\epsilon\sigma\tau_0} + A_2\bar{A}^2\Lambda\Omega^2e^{i\epsilon\sigma\tau_0} + 3A_2A^2\bar{A}\omega_0^2 \end{aligned}$$



**Fig. 17.** Frequency response curve of a micro-plate under electric voltage and superharmonic resonance state for different values of  $V_{DC}$ .

$$+ 2A_2\Lambda^2 A\omega_0^2 - \frac{5A_7A_9^4 V_{DC}^8 A}{A_3^4} - \frac{30A_7A_9^2 V_{DC}^4 A^2 \bar{A}}{A_3^2} - \frac{60A_7A_9^2 V_{DC}^4 \Lambda^2 A}{A_3^2} - \frac{30A_7A_9^2 V_{DC}^4 \bar{A}^2 e^{i\epsilon\sigma\tau_0}}{A_3^2} - 3A_5\bar{A}^2 \Lambda e^{i\epsilon\sigma\tau_0} - 30A_7\bar{A}^2 \Lambda^3 e^{i\epsilon\sigma\tau_0} = 0 \quad (109)$$

Using  $A = \frac{1}{2}a(\tau_2)e^{i\beta(\tau_2)}$  and separating the real and imaginary parts following reduced equations are obtained. To express the obtained equations in their autonomous form one may use the  $\gamma = \sigma\tau_1 - 3\beta$  transformation.

$$\dot{a} = \frac{\sin\gamma}{\omega_0} \times \left( \frac{12A_6A_9V_{DC}^2 a^2 \Lambda}{4A_3} + \frac{2}{4}A_2a^2 \Lambda \omega_0^2 - \frac{20}{16}A_7a^4 \Lambda - \frac{3}{4}A_5a^2 \Lambda - \frac{30}{4}A_7a^2 \Lambda^3 + \frac{1}{4}A_2a^2 \Lambda \Omega^2 \frac{30A_7A_9^2 V_{DC}^4 a^2 \Lambda}{4A_3^2} \right), \quad (110)$$

$$a\dot{\gamma} = a\sigma + \frac{3}{\omega_0} \times \left( -\frac{3A_5a^3}{8} - \frac{A_1A_9V_{DC}^2 a\omega_0^2}{2A_3} + \frac{2A_6A_9^3 V_{DC}^6 a}{A_3^3} + \frac{12A_6A_9V_{DC}^2 a^3}{8A_3} + 2A_2a\Lambda^2 \Omega^2 + \frac{24A_6A_9V_{DC}^2 a\Lambda^2}{2A_3} - \frac{60}{8}A_7a^3 \Lambda^2 - \frac{3A_5A_9^2 V_{DC}^4 a}{2A_3^2} - 3A_5a\Lambda^2 + \frac{A_4A_9V_{DC}^2 a}{A_3} + \frac{A_2A_9^2 V_{DC}^4 a\omega_0^2}{2A_3^2} - \frac{10A_7a^5}{32} + \frac{3A_2a^3 \omega_0^2}{8} - 15A_7a\Lambda^4 + A_2\Lambda^2 a\omega_0^2 - \frac{5A_7A_9^4 V_{DC}^8 a}{2A_3^4} - \frac{30A_7A_9^2 V_{DC}^4 a^3}{8A_3^2} - \frac{60A_7A_9^2 V_{DC}^4 \Lambda^2 a}{2A_3^2} + \left( \frac{12A_6A_9V_{DC}^2 a^2 \Lambda}{4A_3} + \frac{A_2a^2 \Lambda \omega_0^2}{2} - \frac{20}{16}A_7a^4 \Lambda + \frac{1}{4}A_2a^2 \Lambda \Omega^2 \frac{30A_7A_9^2 V_{DC}^4 a^2 \Lambda}{4A_3^2} - \frac{3}{4}A_5a^2 \Lambda - \frac{30}{4}A_7a^2 \Lambda^3 \right) \cos\gamma \right) \quad (111)$$

For steady state, the time derivative terms should be vanished, in Eq. (110) and (111), i.e.,  $\dot{a} = \dot{\gamma} = 0$ . Now by eliminating  $\gamma$  from the above two equations, one can obtain a closed-form equation which can be used for finding the frequency response of the system:

$$\sigma = \frac{-3}{a\omega_0} \times \left( \pm \left( \frac{12A_6A_9V_{DC}^2 a^2 \Lambda}{4A_3} + \frac{A_2a^2 \Lambda \omega_0^2}{2} - \frac{5A_7a^4 \Lambda}{4} - \frac{3A_5a^2 \Lambda}{4} + \frac{A_2a^2 \Lambda \Omega^2}{4} \frac{30A_7A_9^2 V_{DC}^4 a^2 \Lambda}{4A_3^2} - \frac{30A_7a^2 \Lambda^3}{4} \right) - \frac{3A_5a^3}{8} - 3A_5a\Lambda^2 - \frac{A_1A_9V_{DC}^2 a\omega_0^2}{2A_3} + 2A_2a\Lambda^2 \Omega^2 + \frac{2A_6A_9^3 V_{DC}^6 a}{A_3^3} + \frac{12A_6A_9V_{DC}^2 a^3}{8A_3} + \frac{24A_6A_9V_{DC}^2 a\Lambda^2}{2A_3} - \frac{60}{8}A_7a^3 \Lambda^2 - \frac{3A_5A_9^2 V_{DC}^4 a}{2A_3^2} + \frac{A_4A_9V_{DC}^2 a}{A_3} + \frac{A_2A_9^2 V_{DC}^4 a\omega_0^2}{2A_3^2} - 15A_7a\Lambda^4 - \frac{10A_7a^5}{32} + \frac{3A_2a^3 \omega_0^2}{8} \right)$$



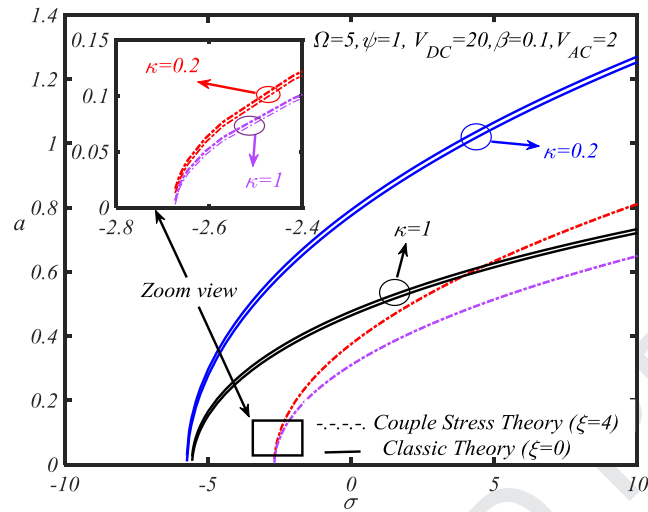


Fig. 18. Frequency Response curve of a micro-plate in subharmonic resonance conditions for various values of  $\kappa$ .

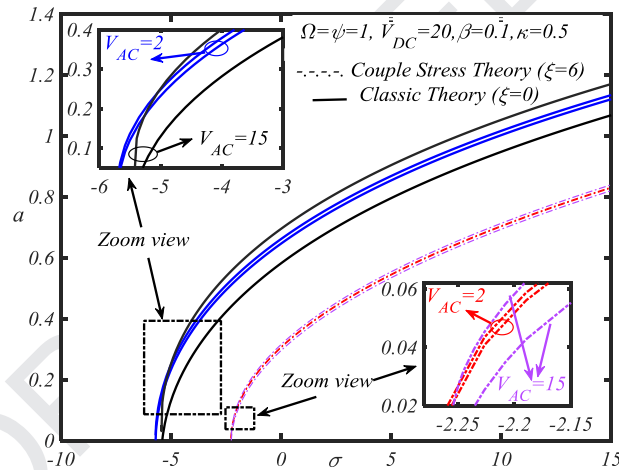


Fig. 19. Frequency response curve for a micro-plate under subharmonic resonance conditions for different  $V_{AC}$  values.

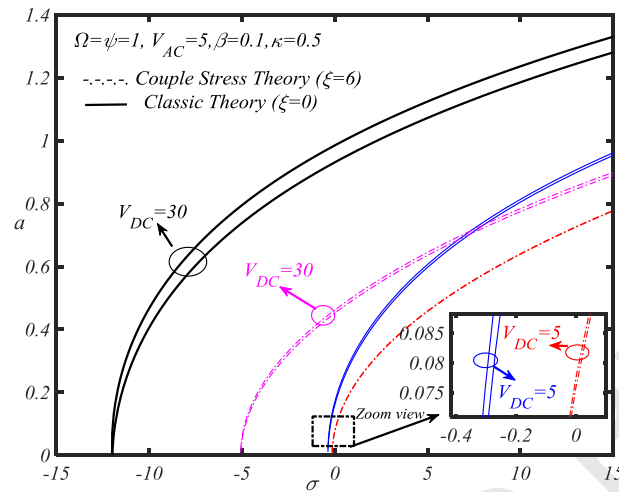
$$+A_2 \Lambda^2 a \omega_0^2 - \frac{5A_7 A_9^4 V_{DC}^8 a}{2A_3^4} - \frac{30A_7 A_9^2 V_{DC}^4 a^3}{8A_3^2} - \frac{60A_7 A_9^2 V_{DC}^4 \Lambda^2 a}{2A_3^2} \quad (112)$$

Above equation shows the system has a trivial state response (i.e.,  $a = 0$ ) and a non-trivial response. The frequency response curve of a micro-plate in subharmonic resonance conditions for various values of  $\kappa$  is shown in Fig. 18. From Fig. 18, it can be seen that by increasing the value of  $\kappa$ , the frequency response curves are shifted to the right side. The frequency response curve of a micro-plate under subharmonic resonance conditions for different  $V_{AC}$  values is shown in Fig. 19. According to this figure, it can be concluded that by increasing the  $V_{AC}$  voltage, the frequency response curves are closer to the vertical axis and the distance between the two branches of each curve will be greater. It is also observed that the increase in  $V_{AC}$  voltage does not affect the slope of the diagrams.

The frequency response curve of a micro-plate under AC voltage  $V_{AC} = 5$  (vot) under subharmonic resonance conditions for different values of  $V_{DC}$  is shown in Fig. 20. As can be seen, increasing the  $V_{DC}$  voltage reduces the stiffness of the micro-plate.

## 6. Conclusions

In this paper, we study the static behavior of a rectangular micro-plate under a constant electrostatic voltage  $V_{DC}$  and its dynamic response when subjected to electrical forces consisting of a constant voltage  $V_{DC}$  or alternating voltage  $V_{AC}$ . Nonlinear von Kármán's relations are used for developing nonlinear governing differential equations of motion for thin micro-plates which are solved by Galerkin and Multiple scale methods. The boundary conditions of the micro-plate have



**Fig. 20.** The frequency response curve for a micro-plate under subharmonic resonance conditions for different  $V_{DC}$  values.

been assumed to be clamped with immovable edges. Numerical results presented in this paper leading to the following findings:

- Increasing  $\kappa$  (gap-to-thickness ratio) leads to an increase in the value of pull-in critical voltage and the maximum pull-in deflection of the micro-plate (Figs. 5 and 10)
- Increasing  $\psi$  (length-to-width ratio) enhances the critical electrostatic pull-in voltage and micro-plate stiffness (Fig. 6)
- It is observed that two types of critical dynamic voltage can occur in clamped micro-plates. In the first type, the micro-plate experiences a sudden fluctuation with a large amplitude (without any contact with the fixed electrode) by reaching the critical voltage. In the second type, it collides the fixed electrode due to the high vibration amplitude. In either case, the system enters an unstable region that should be avoided. (Fig. 12)
- It can be recognized that the critical dynamic voltage of clamped micro-plates is a function of the excitation force frequency for a specific DC voltage.
- It is found that the bending stiffness of the micro-plate decreases in the frequency response curve by increasing the constant DC voltage (Fig. 8). However, an increase in the alternating AC voltage does not affect the stiffness of the micro-plate (Fig. 9)

## Acknowledgements

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## Appendix A: Brief Review on Modified Couple Stress Theory (MCST)

Recently, the importance of vibration at higher mode numbers besides wave propagation in wavelength at the size of the lattice of the medium attracted significant attention. So, the size effect is expected to become remarkable and from a physical point of view, the unusable of classical continuum theory in the above-mentioned cases can be demonstrated, because the wavelength approaches the scale of nanostructures and classical continuum theory fails to anticipate the size dependency [78,79]. To this end, different discrete models or continuum ones have been so far extended. Many attempts have been made to present the non-classical continuum theories by incorporating nonlocality and higher gradient of displacement in kinetic description, such as nonlocal elasticity theory, couple stress theory and strain gradient theory. Recently, studies based on these theories have been an area of active research. Although elegant, contrary to their determinative role, none of them has hitherto been provided the correct physical interpretation of characteristic length scale. Thus, it remains to be determined how the accuracy of these studies can be verified. While, the results of molecular dynamics (MDs) simulations are presented for comparison purposes, there is still controversy in results [79,80].

According to MCST, both strain and curvature tensors contribute to the strain energy density. Based on MCST, the strain energy ( $U$ ) in an isotropic linear elastic material occupying region  $\Omega$  can be written as [81]:

$$U = \frac{1}{2} \int_{\Omega} (\sigma_{ij} \varepsilon_{ij} + m_{ij} \chi_{ij}) d\Omega \quad (A.1)$$

where  $\sigma_{ij}$ ,  $\varepsilon_{ij}$ ,  $m_{ij}$  and  $\chi_{ij}$  are Cauchy stress, strain, deviatoric part of couple stress tensor and symmetric curvature tensors, respectively. The constitutive relations for these tensors are defined as [82]:

$$\sigma_{ij} = \lambda \varepsilon_{ii} + 2\mu \varepsilon_{ij} \quad (\text{A.2})$$

$$\varepsilon_{ij} = \frac{1}{2}(u_{i,j} + u_{j,i}) \quad (\text{A.3})$$

$$m_{ij} = 2\mu l^2 \chi_{ij} \quad (\text{A.4})$$

$$\chi_{ij} = \frac{1}{2}(\theta_{i,j} + \theta_{j,i}) \quad (\text{A.5})$$

where  $u_i$  is the displacement vector,  $\lambda = \frac{E\nu}{(1+\nu)(1-2\nu)}$  and  $\mu = \frac{E}{2(1+\nu)}$  are Lamé's constants ( $\mu$  is also known as shear modulus),  $\theta_i$  is the rotation vector defined as Eq. (A.6) [82] and  $l$  is a material length scale parameter, which can be estimated through experimental tests [83,84], or by simulation technics such as molecular dynamics (MD), which are widely used in monitoring the behavior of a specific system of atoms during dynamic processes [85].

In comparison with other non-classical theories that have two or three parameters as length scales, modified couple stress theory by having only one length scale parameter is one the most appropriate theories in micro scale analysis. This assumption in the framework of couple stress theory might be promising and have potential applications for experimental investigations; determining only one constant via experimental methods is much easier than measuring more constants.

$$\theta_i = \frac{1}{2}e_{ijk}u_{k,j} \quad (\text{A.6})$$

In Eq. (A.6),  $e_{ijk}$  ( $i, j, k = 1, 2, 3$ ) represents the permutation symbol. Compared to the classical continuum mechanics, the Modified couple stress theory has one additional parameter (high-order material length scale) other than two classical Lamé's constants in constitutive equations for isotropic elastic materials.

## Appendix B

Hamilton's principle can be expressed as [86]:

$$\int_{t_1}^{t_2} (\delta T - \delta U + \delta W_{\text{ext}}) dt = 0 \quad (\text{B.1})$$

where  $\delta T$ ,  $\delta U$  and  $\delta W_{\text{ext}}$  denote, virtual kinetic energy, virtual strain energy and virtual work done by external forces, respectively. The virtual kinetic energy is:

$$\begin{aligned} \delta T &= \int_Z \int_A (\rho \dot{D} \cdot \delta D) dA dZ \\ &= - \int_A \int_{-h/2}^{h/2} \rho \left( \frac{\partial U_1}{\partial t} \delta \frac{\partial U_1}{\partial t} + \frac{\partial V_1}{\partial t} \delta \frac{\partial V_1}{\partial t} + \frac{\partial W_1}{\partial t} \delta \frac{\partial W_1}{\partial t} \right) dZ dA \end{aligned} \quad (\text{B.2})$$

In Eq. (B.2) overdot denotes the differentiation with respect to the time variable  $t$ , and  $A$  represents the mid-plane surface and  $\rho$  is the density of micro-plate and:

$$\begin{Bmatrix} I_0 \\ I_1 \\ I_2 \end{Bmatrix} = \int_Z \rho \begin{Bmatrix} 1 \\ Z \\ Z^2 \end{Bmatrix} dZ = \rho \begin{Bmatrix} h \\ 0 \\ \frac{h^3}{12} \end{Bmatrix}, \quad (\text{B.3})$$

$$\delta D = (\delta U - Z \delta W_{,X}) \hat{I} + (\delta V - Z \delta W_{,Y}) \hat{J} + \delta W \hat{K}, \quad (\text{B.4a})$$

$$\ddot{D} = (\ddot{U} - Z \ddot{W}_{,X}) \hat{I} + (\ddot{V} - Z \ddot{W}_{,Y}) \hat{J} + \ddot{W} \hat{K} \quad (\text{B.4b})$$

In equations above,  $D$  represents displacement and is defined as  $D = U_1 \hat{I} + V_1 \hat{J} + W_1 \hat{K}$  and overdot denotes the differentiation concerning the time variable ( $t$ ). In consideration of Eq. (A.1), the expression for virtual strain energy can be expressed as:

$$\begin{aligned} \delta U &= \int_A \int_{-h/2}^{h/2} \{ (\sigma_{ij} + \sigma_{ij}^r) \delta \varepsilon_{ij} + m_{ij} \delta \chi_{ij} \} dZ dA \\ &= \int_A \int_{-h/2}^{h/2} (\sigma_{XX} \delta \varepsilon_{XX} + \sigma_{YY} \delta \varepsilon_{YY} + \sigma_{XY} \delta \gamma_{XY} + \sigma_{XX}^r \delta \varepsilon_{XX} + \sigma_{YY}^r \delta \varepsilon_{YY} \\ &\quad + m_{XX} \delta \chi_{XX} + m_{YY} \delta \chi_{YY} + 2m_{XY} \delta \chi_{XY} + 2m_{XZ} \delta \chi_{XZ} + 2m_{ZY} \delta \chi_{ZY}) dZ dA \end{aligned} \quad (\text{B.5})$$

533 Strain energy can be written by considering Eq. (A.1):

$$\begin{aligned}
 U &= \frac{1}{2} \times \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} \int_{-\frac{h}{2}}^{\frac{h}{2}} \mathfrak{S} dX dY dZ, \\
 \mathfrak{S} &= \left\{ \left( \frac{E}{1-\nu^2} \left( \frac{\partial U}{\partial X} + \frac{1}{2} \left( \frac{\partial W}{\partial X} \right)^2 - ZW_{,xx} + \nu \left( \frac{\partial V}{\partial Y} + \frac{1}{2} \left( \frac{\partial W}{\partial Y} \right)^2 - ZW_{,yy} \right) \right) + \sigma_X^r \right\} \right. \\
 &\quad \times \left( \frac{\partial U}{\partial X} + \frac{1}{2} \left( \frac{\partial W}{\partial X} \right)^2 - ZW_{,xx} \right) + \left( \frac{E}{1-\nu^2} \left( \frac{\partial V}{\partial Y} + \frac{1}{2} \left( \frac{\partial W}{\partial Y} \right)^2 - ZW_{,yy} \right) + \sigma_Y^r \right) \\
 &\quad \times \left( \frac{\partial V}{\partial Y} + \frac{1}{2} \left( \frac{\partial W}{\partial Y} \right)^2 - ZW_{,yy} \right) + \frac{E}{2(1+\nu)} \left( \frac{\partial U}{\partial Y} + \frac{\partial V}{\partial X} + \frac{\partial W}{\partial X} \frac{\partial W}{\partial Y} - 2ZW_{,xy} \right)^2 \\
 &\quad + \frac{2El^2}{1+\nu} \left( \frac{\partial^2 W}{\partial X \partial Y} \right)^2 + \frac{El^2}{2(1+\nu)} \left( \frac{\partial^2 W}{\partial Y^2} - \frac{\partial^2 W}{\partial X^2} \right)^2 + \frac{El^2}{8(1+\nu)} \left( \frac{\partial^2 V}{\partial X^2} - \frac{\partial^2 U}{\partial X \partial Y} \right)^2 \\
 &\quad + \frac{El^2}{8(1+\nu)} \left( \frac{\partial^2 V}{\partial X \partial Y} - \frac{\partial^2 U}{\partial Y^2} \right)^2 \quad (B.6)
 \end{aligned}$$

534 where  $\nu$  and  $E$  are the Poisson's ratio and elastic modulus of the micro-plate, respectively. The virtual work by the dis-  
 535 tributed load can be obtained as [12,13]:

$$\delta W_{\text{ext}} = \int_A f_{\text{external}} \delta W dA = \int_X \int_Y \left( \frac{\varepsilon_0 V(t)^2}{2(g-W)^2} \right) \delta W dX dY \quad (B.7)$$

536 In Eq. (B.7), the external force per unit area of the micro-plate,  $f_{\text{external}}$ , is an electric load composed of a DC component  
 537 ( $V_{\text{DC}}$ ) and an AC component ( $V_{\text{AC}}(t)$ ), and  $\varepsilon_0$  is the permittivity coefficient of vacuum, and  $g$  is the distance between these  
 538 two micro-plates. Resultants of axial residual forces per unit length  $N_{XX}^r$  and  $N_{YY}^r$  are introduced as [87]:

$$N_{XX}^r = \sigma_{XX}^r h, \quad N_{YY}^r = \sigma_{YY}^r h \quad (B.8)$$

539 where  $\sigma_{XX}^r$  and  $\sigma_{YY}^r$  represents the axial residual stresses. Substituting Eqs. (B.2), (B.5) and (B.7) into Eq. (B.1), integrat-  
 540 ing the outcomes by parts, using the fundamental lemma of variational calculus [86] and conducting some mathematical  
 541 manipulations, the following equations of motion:

$$\frac{\partial}{\partial X} Y_{XX} + \frac{\partial}{\partial Y} Y_{XY} + \frac{1}{2} \left( \frac{\partial^2}{\partial X \partial Y} \Gamma_{XZ} + \frac{\partial^2}{\partial Y^2} \Gamma_{YZ} \right) = I_0 \ddot{U}, \quad (B.9a)$$

$$\frac{\partial}{\partial X} Y_{XY} + \frac{\partial}{\partial Y} Y_{YY} - \frac{1}{2} \left( \frac{\partial^2}{\partial X^2} \Gamma_{XZ} + \frac{\partial^2}{\partial X \partial Y} \Gamma_{YZ} \right) = I_0 \ddot{V}, \quad (B.9b)$$

$$\begin{aligned}
 &\frac{\partial^2}{\partial X^2} \Xi_{XX} + 2 \frac{\partial^2}{\partial X \partial Y} \Xi_{XY} + \frac{\partial^2}{\partial Y^2} \Xi_{YY} + Y_{XX} \frac{\partial^2}{\partial X^2} W + 2Y_{XY} \frac{\partial^2}{\partial X \partial Y} W \\
 &\quad + Y_{YY} \frac{\partial^2}{\partial Y^2} W + \frac{\partial Y_{XX}}{\partial X} \frac{\partial W}{\partial X} + \frac{\partial Y_{XY}}{\partial X} \frac{\partial W}{\partial Y} + \frac{\partial Y_{XY}}{\partial Y} \frac{\partial W}{\partial X} + \frac{\partial Y_{YY}}{\partial Y} \frac{\partial W}{\partial Y} \\
 &\quad + N_{XX}^r \frac{\partial^2 W}{\partial X^2} + N_{YY}^r \frac{\partial^2 W}{\partial Y^2} + \frac{\partial^2}{\partial X^2} \Gamma_{XY} - \frac{\partial^2}{\partial Y^2} \Gamma_{XY} + \frac{\partial^2}{\partial X \partial Y} \Gamma_{YY} - \frac{\partial^2}{\partial X \partial Y} \Gamma_{XX} \\
 &\quad + \frac{\varepsilon_0 V(t)^2}{2(g-W)^2} = I_0 \ddot{W} - I_2 \left( \frac{\partial^2 \dot{W}}{\partial X^2} + \frac{\partial^2 \dot{W}}{\partial Y^2} \right) \quad (B.9c)
 \end{aligned}$$

544 and boundary conditions:

$$\left( Y_{XX} + N_{XX}^r + \frac{1}{4} \frac{\partial \Gamma_{XZ}}{\partial Y} \right) n_{XX} + \left( Y_{XY} + \frac{1}{4} \frac{\partial \Gamma_{XZ}}{\partial X} + \frac{1}{2} \frac{\partial \Gamma_{YZ}}{\partial Y} \right) n_{YY} = 0$$

or  $\delta U = 0,$  (B.10a)

$$\frac{1}{4} \Gamma_{XZ} n_{YY} = 0 \quad \text{or} \quad \frac{\partial \delta U}{\partial X} = 0, \quad (B.10b)$$

$$\frac{1}{4} \Gamma_{XZ} n_{XX} + \frac{1}{2} \Gamma_{YZ} n_{YY} = 0 \quad \text{or} \quad \frac{\partial \delta U}{\partial Y} = 0, \quad (B.10c)$$

$$\left( Y_{YY} + N_{YY}^r - \frac{1}{4} \frac{\partial \Gamma_{YZ}}{\partial X} \right) n_{YY} + \left( Y_{XY} - \frac{1}{4} \frac{\partial \Gamma_{YZ}}{\partial Y} - \frac{1}{2} \frac{\partial \Gamma_{XZ}}{\partial X} \right) n_{XX} = 0 \quad \text{or } \delta V = 0, \quad (\text{B.10d})$$

$$\frac{1}{4} \Gamma_{YZ} n_{YY} + \frac{1}{2} \Gamma_{XZ} n_{XX} = 0 \quad \text{or } \frac{\partial \delta V}{\partial X} = 0, \quad (\text{B.10e})$$

$$\frac{1}{4} \Gamma_{YZ} n_{XX} = 0 \quad \text{or } \frac{\partial \delta V}{\partial Y} = 0, \quad (\text{B.10f})$$

$$\begin{aligned} & \left( (Y_{XX} + N_{XX}^r) \frac{\partial W}{\partial X} + (Y_{XY}) \frac{\partial W}{\partial Y} + \frac{\partial \Xi_{XX}}{\partial X} + \frac{\partial \Xi_{XY}}{\partial Y} \right. \\ & \quad \left. - \frac{1}{2} \frac{\partial}{\partial Y} \Gamma_{XX} + \frac{\partial}{\partial X} \Gamma_{XY} + \frac{1}{2} \frac{\partial}{\partial Y} \Gamma_{YY} + I_2 \frac{\partial \ddot{W}}{\partial X} \right) n_{XX} \\ & + \left( (Y_{YY} + N_{YY}^r) \frac{\partial W}{\partial Y} + (Y_{XY}) \frac{\partial W}{\partial X} + \frac{\partial \Xi_{YY}}{\partial Y} + \frac{\partial \Xi_{XY}}{\partial X} \right. \\ & \quad \left. - \frac{1}{2} \frac{\partial}{\partial X} \Gamma_{XX} - \frac{\partial}{\partial Y} \Gamma_{XY} + \frac{1}{2} \frac{\partial}{\partial X} \Gamma_{YY} + I_2 \frac{\partial \ddot{W}}{\partial Y} \right) n_{YY} = 0 \\ & \quad \text{or } \delta W = 0, \end{aligned} \quad (\text{B.10g})$$

$$\begin{aligned} & (\Xi_{XX} + \Gamma_{XY}) n_{XX} + \left( \Xi_{XY} - \frac{1}{2} \Gamma_{XX} + \frac{1}{2} \Gamma_{YY} \right) n_{YY} = 0 \\ & \quad \text{or } \frac{\partial \delta W}{\partial X} = 0, \end{aligned} \quad (\text{B.10h})$$

$$\begin{aligned} & (\Xi_{YY} - \Gamma_{XY}) n_{YY} + \left( \Xi_{XY} - \frac{1}{2} \Gamma_{XX} + \frac{1}{2} \Gamma_{YY} \right) n_{XX} = 0 \\ & \quad \text{or } \frac{\partial \delta W}{\partial Y} = 0 \end{aligned} \quad (\text{B.10i})$$

are derived. In above equations  $n_i$  ( $i = XX, YY$ ) are the components of a normal vector to the boundary of the mid-plane of micro-plate. To extract the governing equations of motion in terms of displacements, stress and couple stress consequents can be written as:

$$\begin{Bmatrix} \Xi_{XX} \\ \Xi_{YY} \\ \Xi_{XY} \end{Bmatrix} = \int_{-h/2}^{h/2} Z \begin{Bmatrix} \sigma_{XX} \\ \sigma_{YY} \\ \sigma_{XY} \end{Bmatrix} dZ = \frac{-Eh^3}{12(1-\nu^2)} \begin{bmatrix} 1 & \nu & 0 \\ \nu & 1 & 0 \\ 0 & 0 & (1-\nu)/2 \end{bmatrix} \begin{Bmatrix} \frac{\partial^2 W}{\partial X^2} \\ \frac{\partial^2 W}{\partial Y^2} \\ 2 \frac{\partial^2 W}{\partial X \partial Y} \end{Bmatrix} \quad (\text{B.11a})$$

$$\begin{Bmatrix} \Gamma_{XX} \\ \Gamma_{YY} \\ \Gamma_{XY} \\ \Gamma_{XZ} \\ \Gamma_{YZ} \end{Bmatrix} = \int_{-h/2}^{h/2} \begin{Bmatrix} m_{XX} \\ m_{YY} \\ m_{XY} \\ m_{XZ} \\ m_{YZ} \end{Bmatrix} dZ = \frac{Ehl^2}{1+\nu} \begin{bmatrix} 1 & 0 & 0 & 0 & 0 \\ 0 & -1 & 0 & 0 & 0 \\ 0 & 0 & -\frac{1}{2} & 0 & 0 \\ 0 & 0 & 0 & -\frac{1}{4} & 0 \\ 0 & 0 & 0 & 0 & -\frac{1}{4} \end{bmatrix} \begin{Bmatrix} \frac{\partial^2 W}{\partial X \partial Y} \\ \frac{\partial^2 W}{\partial X \partial Y} \\ \frac{\partial^2 W}{\partial X^2} - \frac{\partial^2 W}{\partial Y^2} \\ \frac{\partial^2 U}{\partial X \partial Y} - \frac{\partial^2 V}{\partial X^2} \\ \frac{\partial^2 U}{\partial Y^2} - \frac{\partial^2 V}{\partial X \partial Y} \end{Bmatrix} \quad (\text{B.11b})$$

$$\begin{Bmatrix} Y_{XX} \\ Y_{YY} \\ Y_{XY} \end{Bmatrix} = \int_Z \begin{Bmatrix} \sigma_{XX} \\ \sigma_{YY} \\ \sigma_{XY} \end{Bmatrix} dZ = \frac{Eh}{(1-\nu^2)} \begin{bmatrix} 1 & \nu & 0 \\ \nu & 1 & 0 \\ 0 & 0 & (1-\nu)/2 \end{bmatrix} \begin{Bmatrix} \frac{\partial U}{\partial X} + \frac{1}{2} \left( \frac{\partial W}{\partial X} \right)^2 \\ \frac{\partial V}{\partial Y} + \frac{1}{2} \left( \frac{\partial W}{\partial Y} \right)^2 \\ \frac{\partial U}{\partial Y} + \frac{\partial V}{\partial X} + \frac{\partial W}{\partial X} \frac{\partial W}{\partial Y} \end{Bmatrix} \quad (\text{B.11c})$$

## Appendix C

Constant  $C_1$ ,  $C_2$ ,  $A_n$ , and  $B_n$ , introduced in Eq. (36), are determined here. On the determination of these constants, it is observed that:

$$[F_{c,XY}]_{X=\pm a/2} = Eg^2 W_m^2 \sum_{n=1}^n \frac{\mp 2An\pi \sin(n\pi)\lambda}{na(U\lambda P + n\pi)} \left( -2 \frac{\pi Un(2\pi nYQ + aR)}{a^2} \right)$$

$$\begin{aligned}
& + 2 \frac{n\pi (P\pi n + U\lambda)R}{a\lambda} \Big) - \frac{2BTn\pi}{\lambda^2 nb (\sinh(n\pi\lambda) \cosh(n\pi\lambda) + n\pi\lambda)} \\
& \times \left( \pm 2 \frac{(\sinh(n\pi\lambda) + n\pi\lambda \cosh(n\pi\lambda))n\pi \sinh(K)}{b} + \right. \\
& \left. \frac{(\mp 2 \sinh(n\pi\lambda) \cosh(K)K \mp 2 \sinh(n\pi\lambda) \sinh(K))n\pi}{b} \right), \\
[F_{c,XY}]_{Y=\pm b/2} &= \sum_{n=1}^n \pm 4 \frac{Eg^2 W_m^2 B n n \pi^3 \sin(n\pi)}{b^3 \lambda^2 (\sinh(n\pi\lambda) \cosh(n\pi\lambda) + n\pi\lambda)} \\
& \times \left( -\cosh(n\pi\lambda) \sinh\left(2 \frac{n\pi X}{b}\right) b\lambda + 2 \sinh(n\pi\lambda) HX \right) \\
& - \frac{2An\pi\lambda J}{na(U\lambda P + n\pi)} \left( \pm 2 \frac{n\pi \sinh(S)(P\pi n + U\lambda)}{a\lambda} \mp \right. \\
& \left. 2 \frac{\pi Un(\pi \cosh(S)bn + \sinh(S)a)}{a^2} \right) \tag{C.1}
\end{aligned}$$

561 where

$$\begin{aligned}
H &= \cosh\left(2 \frac{n\pi X}{b}\right), S = \frac{n\pi b}{a}, T = \sin\left(2 \frac{n\pi Y}{b}\right), P = \cosh\left(\frac{n\pi}{\lambda}\right), \\
U &= \sinh\left(\frac{n\pi}{\lambda}\right), K = \frac{n\pi a}{b}, Q = \cosh\left(2 \frac{n\pi Y}{a}\right), R = \sinh\left(2 \frac{n\pi Y}{a}\right) \tag{C.2}
\end{aligned}$$

562 Introducing equation  $F = F_c + F_p$  into the equation of boundary conditions (18) and taking into account Eqs. (35) and (C.2),  
563 the constants  $C_1$  and  $C_2$  are determined.

## 564 Appendix D

565 The coefficients  $A_i$ ,  $i = 1, 2, \dots, 8$ , of Eq. (56) are presented as:

$$\begin{aligned}
A_1 &= -\frac{25}{18}, A_2 = \frac{1225}{2304}, A_3 = 16/3 \left( \frac{(\psi^4 + 2/3\psi^2 + 1)(\xi + 1)\pi^2}{+ 1/4\psi^2 N_y + N_x/4} \right) \pi^2, \\
A_4 &= -\frac{80\pi^2 ((\psi^4 + 4/5\psi^2 + 1)(\xi + 1)\pi^2 + 1/4\psi^2 N_y + N_x/4)}{9}, \\
A_8 &= -\frac{16\beta (V_{DC} + V_{AC} \cos(\Omega t))^2}{9}, \\
A_6 &= \frac{(\nu + 1)\pi^4 (264\psi^{16}\nu - 504\psi^{16} + 2772\psi^{14}\nu - 5532\psi^{14} + 11267\psi^{12}\nu \\
& - 22922\psi^{12} + 24218\psi^{10}\nu - 47063\psi^{10} + 34158\psi^8\nu - 57003\psi^8 \\
& + 24218\psi^6\nu - 35873\psi^6 + 11267\psi^4\nu - 14027\psi^4 + 2772\psi^2\nu \\
& - 3012\psi^2 + 264\nu - 264)\kappa^2}{96(\psi^2 + 1/4)^2(\psi^2 + 1)^2(\psi^2 + 4)^2}, \tag{D.1}
\end{aligned}$$

566

$$\begin{aligned}
A_7 &= -\frac{(21\nu + 21)\pi^4 (480\psi^{16}\nu - 880\psi^{16} + 5040\psi^{14}\nu - 9640\psi^{14} \\
& + 20846\psi^{12}\nu - 40271\psi^{12} + 45740\psi^{10}\nu - 83815\psi^{10} + 65388\psi^8\nu \\
& - 103463\psi^8 + 45740\psi^6\nu - 65165\psi^6 + 20846\psi^4\nu - 25446\psi^4 + \\
& 5040\psi^2\nu - 5440\psi^2 + 480\nu - 480)\kappa^2}{8192(\psi^2 + 1/4)^2(\psi^2 + 1)^2(\psi^2 + 4)^2}
\end{aligned}$$



567

$$A_5 = \frac{175\pi^2 \times \left\{ \left( \left( \frac{4}{5} - \frac{68v^2}{175} + \frac{72v}{175} \right) \kappa^2 + \xi + 1 \right) \pi^2 \psi^{16} + \left( \left( \frac{68}{175} - \frac{68v^2}{175} \right) \kappa^2 + \xi + 1 \right) \pi^2 \right.}{48(\psi^2 + 4)^2(\psi^2 + 1)^2(\psi^2 + 1/4)^2}$$

568 It is worth mentioning that, we assume in the abovementioned equations that voltage varies according to  $V(t) = V_{DC} +$   
569  $V_{AC} \cos(\Omega t)$ ,  $\Omega$  is the excitation frequency

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